

# Chapter 5 Reflection and Transmission

## 5.1 Introduction

- moving on from unbounded media to 2 semi-infinite media divided by a planar boundary
- Key parameters:
  - \* reflection & transmission coefficients
  - \* Brewster angle (no reflection)
  - \* Critical angle (100% reflection)

## 5.2 Normal Incidence - Lossless Media

→ see Fig 5-1

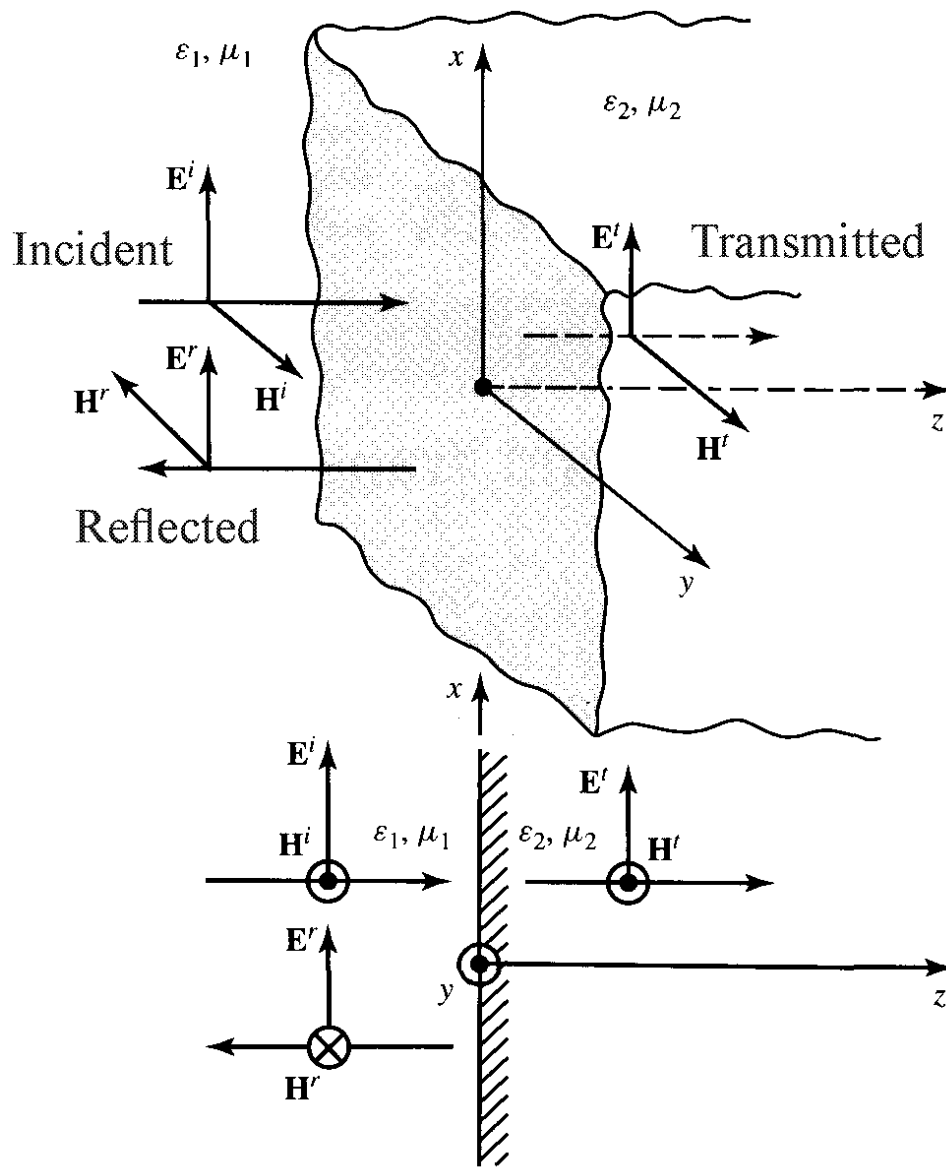
We'll assume the electric field(s) are in the x-direction

$$\vec{E}^i = \hat{a}_x E_0 e^{-j\beta_1 z} \quad \text{incident} \quad z \leq 0$$

$$\vec{E}^r = \hat{a}_x \Gamma^b E_0 e^{+j\beta_1 z} \quad \text{reflected} \quad z \leq 0$$

$$\vec{E}^t = \hat{a}_x T^b E_0 e^{-j\beta_2 z} \quad \text{transmitted} \quad z \geq 0$$

where  $\Gamma^b$  &  $T^b$  are the reflection and transmission coefficients @ the boundary ( $z=0$ ). Note that all the electric fields are assumed to be oriented the same.



**Figure 5-1** Wave reflection and transmission at normal incidence by a planar interface.

*Advanced Engineering Electromagnetics* (Second Edition), Balanis, Wiley, 2012, ISBN-10: 0470589485, ISBN-13: 978-0470589489.

S.2 cont.

Using  $\bar{H} = \frac{\hat{a}_k \times \bar{E}}{\eta}$  or Faraday's Law, we can find the corresponding magnetic fields

$$\bar{H}^i = \frac{+\hat{a}_z \times \bar{E}^i}{\eta_1} = \hat{a}_y \frac{E_0}{\eta_1} e^{-j\beta_1 z} \quad z \leq 0$$

$$\bar{H}^r = \frac{-\hat{a}_z \times \bar{E}^r}{\eta_1} = -\hat{a}_y \frac{\Gamma^b E_0}{\eta_1} e^{j\beta_1 z} \quad z \leq 0$$

$$\bar{H}^t = \frac{+\hat{a}_z \times \bar{E}^t}{\eta_2} = \hat{a}_y \frac{T^b E_0}{\eta_2} e^{-j\beta_2 z} \quad z \geq 0$$

Assuming no sources on the boundary, we apply the tangential boundary conditions

@  $z = 0$

$$\bar{E}^i(z=0) + \bar{E}^r(z=0) = \bar{E}^t(z=0)$$

$$\bar{H}^i(z=0) + \bar{H}^r(z=0) = \bar{H}^t(z=0)$$

$\Downarrow$

$$E_0 + E_0 \Gamma^b = E_0 T^b \rightarrow 1 + \Gamma^b = T^b$$

$$\frac{E_0}{\eta_1} - \frac{\Gamma^b E_0}{\eta_1} = \frac{T^b E_0}{\eta_2} \rightarrow \frac{1}{\eta_1} (1 - \Gamma^b) = \frac{T^b}{\eta_2}$$

Solve 2 eq's - 2 unknowns  $\Downarrow$

S.2 cont.

$$\Gamma^b = \frac{E^r}{E^i} = \frac{-H^r}{H^i} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1}$$

$$T^b = \frac{E^t}{E^i} = \frac{\eta_2}{\eta_1} \frac{H^t}{H^i} = 1 + \Gamma^b = \frac{2\eta_2}{\eta_2 + \eta_1}$$

Note:  $\Gamma^b = 0$  only if  $\eta_1 = \eta_2 \Rightarrow \epsilon_2 = \epsilon_1$  for non-magnetic materials

We can define reflection & transmission coefficients away from the boundary @ locations  $z = -l_1$  and  $z = l_2$  as

$$\Gamma(z = -l_1) = \frac{E^r(z)}{E^i(z)} \Big|_{z=-l_1} = \frac{\Gamma_b E_0 e^{-j\beta_1 l_1}}{E_0 e^{+j\beta_1 l_1}}$$

$$\underline{\Gamma(z = -l_1) = \Gamma_b e^{-j2\beta_1 l_1}}$$

$$T(z_2 = l_2, z_1 = -l_1) = \frac{E^t(z_2 = l_2)}{E^i(z_1 = -l_1)} = \frac{T_b E_0 e^{-j\beta_2 l_2}}{E_0 e^{+j\beta_1 l_1}}$$

$$\underline{T(z_2 = l_2, z_1 = -l_1) = T_b e^{-j(\beta_2 l_2 + \beta_1 l_1)}}$$

Note:  $l_1 + l_2 > 0$

S.2 cont.

What about power flow? The time-ave Poynting vectors are:

$$\underline{\bar{S}_{ave}^i} = \frac{1}{2} \operatorname{Re}(\bar{E}^i \times \bar{H}^{i*}) = \underline{\hat{a}_z \frac{|E_0|^2}{2\eta_1}} \quad z < 0$$

$$\begin{aligned} \bar{S}_{ave}^r &= \frac{1}{2} \operatorname{Re}(\bar{E}^r \times \bar{H}^{r*}) = -\hat{a}_z |\Gamma^b|^2 \frac{|E_0|^2}{2\eta_1} \\ &= \underline{-\hat{a}_z |\Gamma^b|^2 |\bar{S}_{ave}^i|} \quad z < 0 \end{aligned}$$

$$\begin{aligned} \bar{S}_{ave}^t &= \frac{1}{2} \operatorname{Re}(\bar{E}^t \times \bar{H}^{t*}) = \hat{a}_z |T^b|^2 \frac{|E_0|^2}{2\eta_2} \\ &= \hat{a}_z |T^b|^2 \frac{\eta_1}{\eta_2} \frac{|E_0|^2}{2\eta_1} \quad \left. \begin{array}{l} \text{massage} \\ \text{to get} \\ |\bar{S}_{ave}^i| \end{array} \right\} \\ &= \hat{a}_z |T^b|^2 \frac{\eta_1}{\eta_2} |\bar{S}_{ave}^i| \quad \text{not obvious} \\ &= \underline{\hat{a}_z (1 - |\Gamma^b|^2) |\bar{S}_{ave}^i|} \quad \text{makes sense} \\ & \quad \quad \quad z > 0 \end{aligned}$$

Comparing the underlined equations, we see that power is conserved.

⇒ In media 1, a standing wave will exist

$$\begin{aligned} \bar{E}^1 &= \bar{E}^i + \bar{E}^r = \hat{a}_x E_0 e^{-j\beta_1 z} (1 + \Gamma^b e^{j2\beta_1 z}) \\ &= \hat{a}_x E_0 e^{-j\beta_1 z} (1 + \Gamma(z)) \quad z < 0 \end{aligned}$$

S.2 cont.

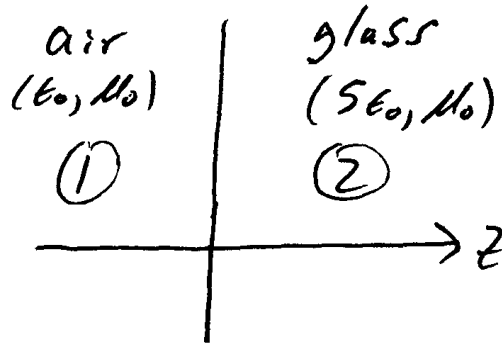
Comparing the max & min values of  $\bar{E}'$ , as done in Chap 4, leads to the standing wave ratio:

$$\underline{SWR = \frac{|\bar{E}'|_{\max}}{|\bar{E}'|_{\min}} = \frac{1 + |\Gamma^b|}{1 - |\Gamma^b|}}$$

For non-magnetic media (e.g.,  $\mu_1 = \mu_2 = \mu_0$ ) or media w/ identical magnetic properties, the SWR simplifies to

$$\underline{SWR = \begin{cases} \sqrt{\frac{\epsilon_1}{\epsilon_2}} & \epsilon_1 > \epsilon_2 \\ \sqrt{\frac{\epsilon_2}{\epsilon_1}} & \epsilon_2 > \epsilon_1 \end{cases}}$$

**Example-** A UPW in air ( $\epsilon_0, \mu_0, z < 0$ ) is normally incident on a glass half-space ( $5\epsilon_0, \mu_0, z > 0$ ). The 2.4 GHz incident electric field is oriented in the  $x$ -direction and has a field strength of 0.6 V/m at  $z = 0$ . Analyze and determine the various associated fields, power densities, and other related quantities.



$$\beta_1 = \omega \sqrt{\mu_1 \epsilon_1} = 2\pi (2.4 \times 10^9) \sqrt{4\pi \times 10^{-7} (8.8541878 \times 10^{-12})}$$

$$= \underline{50.3003 \text{ rad/m}}$$

$$\eta_1 = \eta_0 = \underline{376.7303 \Omega}$$

$$\beta_2 = \omega \sqrt{\mu_2 \epsilon_2} = 2\pi (2.4 \times 10^9) \sqrt{4\pi \times 10^{-7} (5) (8.8541878 \times 10^{-12})}$$

$$= \underline{112.47485 \text{ rad/m}}$$

$$\eta_2 = \sqrt{\frac{\mu_2}{\epsilon_2}} = \frac{\eta_0}{\sqrt{5}} = \frac{376.7303}{\sqrt{5}} = \underline{168.4789 \Omega}$$

$$\Gamma^b = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{168.48 - 376.73}{168.48 + 376.73} = \underline{-0.381966}$$

$$T^b = 1 + \Gamma^b = 1 - 0.381966 = \underline{0.618034}$$

ex. cont.

$$\underline{\bar{E}^i = \hat{a}_x 0.6 e^{-j50.3z} \text{ (V/m)} \quad z \leq 0 \text{ (air)}}$$

$$\bar{E}^r = \hat{a}_x (-0.382) 0.6 e^{+j50.3z}$$

$$\underline{\bar{E}^r = -\hat{a}_x 0.2292 e^{+j50.3z} \text{ (V/m)} \quad z \leq 0 \text{ (air)}}$$

$$\underline{\bar{E}_{air} = \bar{E}^i + \bar{E}^r = \hat{a}_x \left[ 0.6 e^{-j50.3z} - 0.2292 e^{j50.3z} \right] \text{ (V/m)} \quad z \leq 0}$$

$$\bar{E}^t = \hat{a}_x (0.618) 0.6 e^{-j112.475z}$$

$$\underline{\bar{E}_{glass} = \bar{E}^t = \hat{a}_x 0.3708 e^{-j112.475z} \text{ (V/m)} \quad z \geq 0 \text{ (glass)}}$$

$$\underline{\bar{H}^i = \hat{a}_y \frac{0.6}{376.73} e^{-j50.3z} = \hat{a}_y 1.59265 e^{-j50.3z} \text{ (mA/m)} \quad z \leq 0 \text{ (air)}}$$

$$\bar{H}^r = -\hat{a}_y \frac{(-0.382) 0.6}{376.73} e^{+j50.3z}$$

$$\underline{\bar{H}^r = \hat{a}_y 0.60834 e^{+j50.3z} \text{ (mA/m)} \quad z \leq 0 \text{ (air)}}$$

$$\underline{\bar{H}_{air} = \bar{H}^i + \bar{H}^r = \hat{a}_y \left[ 1.593 e^{-j50.3z} + 0.608 e^{j50.3z} \right] \text{ (mA/m)} \quad z \leq 0 \text{ (air)}}$$

$$\bar{H}^t = \hat{a}_y \frac{0.618(0.6)}{168.479} e^{-j112.475z}$$

$$\underline{\bar{H}_{glass} = \bar{H}^t = \hat{a}_y 2.201 e^{-j112.475z} \text{ (mA/m)} \quad z \geq 0 \text{ (glass)}}$$



ex. cont.

$$SWR_{air} = \frac{1 + |-0.382|}{1 - |-0.382|} = \underline{\underline{2.236}}$$

$$\bar{S}_{ave}^i = \frac{1}{2} \text{Re}(\hat{a}_x 0.6 e^{-j50.3z} \times \hat{a}_y 1.59265 \times 10^{-3} e^{+j50.3z})$$

$$\bar{S}_{ave}^i = \hat{a}_z 0.4778 \frac{\text{mW}}{\text{m}^2} \quad z \leq 0 \text{ (air)}$$


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$$\bar{S}_{ave}^r = \frac{1}{2} \text{Re}(-\hat{a}_x 0.2292 e^{j50.3z} \times \hat{a}_y 6.0834 \times 10^{-4} e^{-j50.3z})$$

$$\bar{S}_{ave}^r = -\hat{a}_z 0.0697 \frac{\text{mW}}{\text{m}^2} \quad z \leq 0 \text{ (air)}$$


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$$\bar{S}_{ave}^t = \frac{1}{2} \text{Re}(\hat{a}_x 0.3708 e^{-j112.475z} \times \hat{a}_y 2.201 \times 10^{-3} e^{+j112.475z})$$

$$\bar{S}_{ave}^t = \hat{a}_z 0.4081 \frac{\text{mW}}{\text{m}^2} \quad z \geq 0 \text{ (glass)}$$


---

Check  $\bar{S}_{ave}^i + \bar{S}_{ave}^r \stackrel{?}{=} \bar{S}_{ave}^t$

$$\hat{a}_z (0.4778 - 0.0697) \stackrel{?}{=} \hat{a}_z 0.4081$$

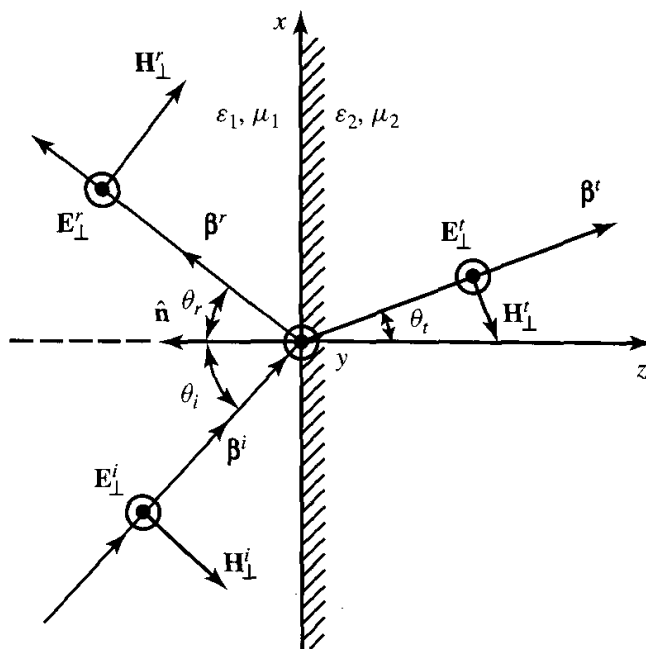
$$\frac{\hat{a}_z 0.4081 \frac{\text{mW}}{\text{m}^2}}{\text{air}} = \frac{\hat{a}_z 0.4081 \frac{\text{mW}}{\text{m}^2}}{\text{glass}} \quad \therefore$$

### 5.3 Oblique Incidence - Lossless Media

- Again, we'll place the boundary @  $z=0$  ( $x$ - $y$  plane)
- Now, the UPW will be incident at an oblique angle
- We'll define the 'plane of incidence' as the plane defined by the surface normal to the boundary (i.e.,  $+\hat{a}_z$ ) and the vector in the direction of incidence ( $x$ - &  $z$ -directions). In this case, the plane of incidence is the  $x$ - $z$  plane.

#### Two cases

- 1) Electric field is orthogonal/perpendicular to plane of incidence (i.e.  $\bar{y}$ -direction)
  - ⇒ Perpendicular/horizontal / E polarization  $\bar{E}_\perp$
- 2) Magnetic field is orthogonal to plane of incidence ⇒ electric is parallel  $\bar{E}_\parallel$ 
  - ⇒ Parallel/vertical / H polarization



**Figure 5-2** Perpendicular (horizontal) polarized uniform plane wave incident at an oblique angle on an interface.

*Advanced Engineering Electromagnetics* (Second Edition), Balanis, Wiley, 2012, ISBN-10: 0470589485, ISBN-13: 978-0470589489.

### 5.3.1 Perpendicular (Horizontal or E) Polarization

$$\bar{E}_\perp^i = \hat{a}_y E_\perp^i e^{-j\beta_i \cdot \bar{r}} = \hat{a}_y E_0 e^{-j\beta_i(x \sin \theta_i + z \cos \theta_i)}$$

$$\begin{aligned} \bar{H}_\perp^i &= (-\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i) H_\perp^i e^{-j\beta_i \cdot \bar{r}} \\ &= (-\hat{a}_x \cos \theta_i + \hat{a}_z \sin \theta_i) \frac{E_0}{\eta_1} e^{-j\beta_i(x \sin \theta_i + z \cos \theta_i)} \end{aligned}$$

⇒ Similar to section 4.2.2 where we defined a  $\bar{\beta}^+ = \beta (\hat{a}_x \sin \theta_i + \hat{a}_z \cos \theta_i)$

The reflected electric field will be in the y-direction ( $\pm$  determined by a new reflection coefficient,

$$\bar{E}_\perp^r = \hat{a}_y E_\perp^r e^{-j\beta_r \cdot \bar{r}} = \hat{a}_y \Gamma_\perp^b E_0 e^{-j\beta_i(x \sin \theta_r - z \cos \theta_r)}$$

$$\begin{aligned} \bar{H}_\perp^r &= (\hat{a}_x \cos \theta_r + \hat{a}_z \sin \theta_r) H_\perp^r e^{-j\beta_r \cdot \bar{r}} \\ &= (\hat{a}_x \cos \theta_r + \hat{a}_z \sin \theta_r) \frac{\Gamma_\perp^b E_0}{\eta_1} e^{-j\beta_i(x \sin \theta_r - z \cos \theta_r)} \end{aligned}$$

Note:  $E_\perp^r = \Gamma_\perp^b E_\perp^i = \Gamma_\perp^b E_0 \Rightarrow \Gamma_\perp^b = \frac{E_\perp^r}{E_\perp^i}$

### S.3.1 cont.

The transmitted fields are given as

$$\vec{E}_\perp^t = \hat{a}_y E_\perp^t e^{-j\vec{\beta}^t \cdot \vec{r}} = \hat{a}_y T_\perp^b E_0 e^{-j\beta_2(x \sin \theta_t + z \cos \theta_t)}$$

$$\begin{aligned} \vec{H}_\perp^t &= (-\hat{a}_x \cos \theta_t + \hat{a}_z \sin \theta_t) H_\perp^t e^{-j\vec{\beta}^t \cdot \vec{r}} \\ &= (-\hat{a}_x \cos \theta_t + \hat{a}_z \sin \theta_t) \frac{T_\perp^b E_0}{\eta_2} e^{-j\beta_2(x \sin \theta_t + z \cos \theta_t)} \end{aligned}$$

Note:  $E_\perp^t = T_\perp^b E_\perp^i = T_\perp^b E_0 \Rightarrow T_\perp^b = \frac{E_\perp^t}{E_\perp^i}$

Next, use our tangential boundary conditions (BCs) to find  $\Gamma_\perp^b$  &  $T_\perp^b$ . First, apply tangential electric BC

$$\vec{E}_{1,t}^i = (\vec{E}_\perp^i + \vec{E}_\perp^r) \Big|_{z=0} = (\vec{E}_\perp^t) \Big|_{z=0} = \vec{E}_{2,t} \quad \begin{array}{l} y\text{-direction} \\ \text{is tangential} \end{array}$$

$$\hat{a}_y E_0 e^{-j\beta_1 x \sin \theta_i} + \hat{a}_y \Gamma_\perp^b E_0 e^{-j\beta_1 x \sin \theta_r} = \hat{a}_y T_\perp^b E_0 e^{-j\beta_2 x \sin \theta_t}$$

$$e^{-j\beta_1 x \sin \theta_i} + \Gamma_\perp^b e^{-j\beta_1 x \sin \theta_r} = T_\perp^b e^{-j\beta_2 x \sin \theta_t} \quad (5-14a)$$

Next, apply tangential magnetic BC (keep  $\hat{a}_x$  terms)

$$\vec{H}_{1,t}^i = (\vec{H}_\perp^i + \vec{H}_\perp^r) \Big|_{z=0} = \vec{H}_\perp^t \Big|_{z=0} = \vec{H}_{2,t} \quad \hat{a}_z \text{ terms are normal}$$

$$\begin{aligned} -\hat{a}_x \frac{E_0}{\eta_1} \cos \theta_i e^{-j\beta_1 x \sin \theta_i} + \hat{a}_x \frac{\Gamma_\perp^b E_0}{\eta_1} \cos \theta_r e^{-j\beta_1 x \sin \theta_r} \\ = -\hat{a}_x \frac{T_\perp^b E_0}{\eta_2} \cos \theta_t e^{-j\beta_2 x \sin \theta_t} \end{aligned}$$

5.3.1 cont.

which simplifies to

$$\frac{1}{\eta_1} \left[ -\cos\theta_i e^{-j\beta_1 x \sin\theta_i} + \Gamma_{\perp}^b \cos\theta_r e^{-j\beta_1 x \sin\theta_r} \right] = \frac{-T_{\perp}^b}{\eta_2} \cos\theta_t e^{-j\beta_2 x \sin\theta_t} \quad (5-14b)$$

Now (5-14a) can only hold true for all values

of  $x$  if  $\sin\theta_i = \sin\theta_r \Rightarrow \underline{\theta_i = \theta_r}$  Snell's Law  
of reflection

$\Rightarrow$  incident & reflected wave must be 'in-phase' @  $z=0$  for all  $x$

This, in turn, yields

$$(1 + \Gamma_{\perp}^b) e^{-j\beta_1 x \sin\theta_i} = T_{\perp}^b e^{-j\beta_2 x \sin\theta_t}$$

which can only hold true for all values  
of  $x$  if

$$\underline{\beta_1 \sin\theta_i = \beta_2 \sin\theta_t} \quad \text{Snell's Law of refraction}$$

With these two relationships, (5-14a) & (5-14b) have identical exponential terms that can be divided out to yield

$$1 + \Gamma_{\perp}^b = T_{\perp}^b$$

$$\frac{\cos\theta_i}{\eta_1} (-1 + \Gamma_{\perp}^b) = -\frac{\cos\theta_t}{\eta_2} T_{\perp}^b$$

2 eqns - 2 unknowns

S.3.1 cont.

Solving  
Fresnel  
reflection  
&  
transmission  
coefficients  
(perpend. pol.)

$$\Gamma_{\perp}^b = \frac{E_{\perp}^r}{E_{\perp}^i} = \frac{\eta_2 \cos \theta_i - \eta_1 \cos \theta_t}{\eta_2 \cos \theta_i + \eta_1 \cos \theta_t}$$

$$T_{\perp}^b = \frac{E_{\perp}^t}{E_{\perp}^i} = \frac{2 \eta_2 \cos \theta_i}{\eta_2 \cos \theta_i + \eta_1 \cos \theta_t}$$

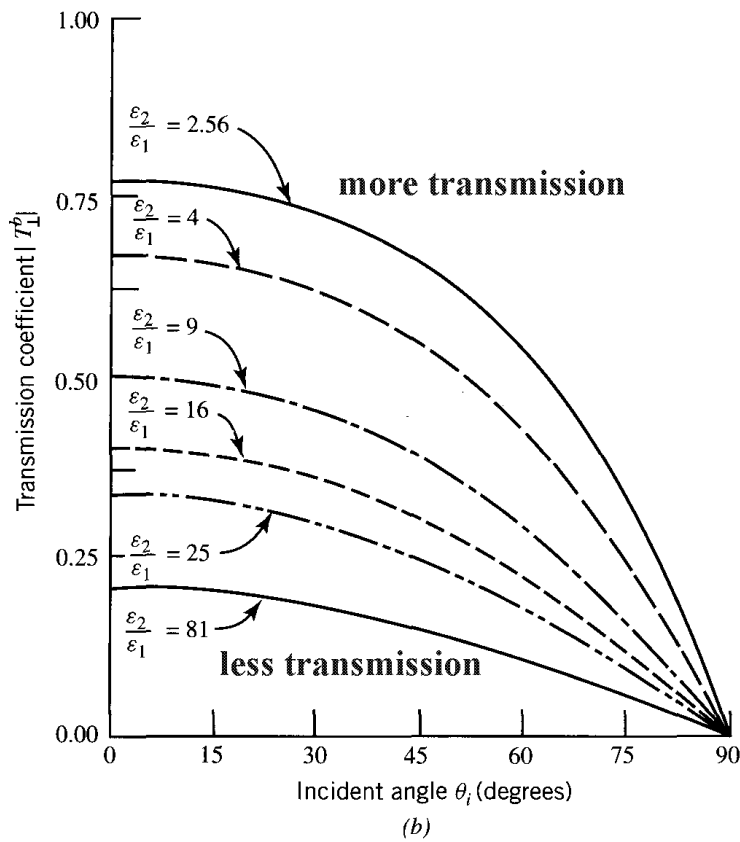
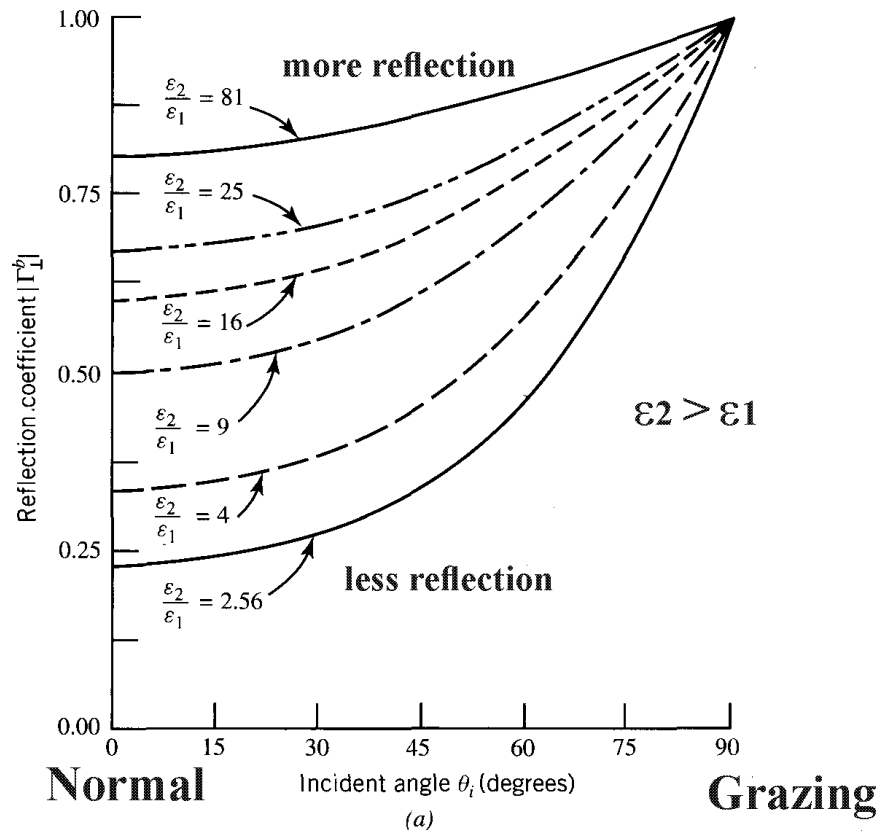
For dielectric media where  $\mu_1 = \mu_2 = \mu_0$ ,  
these become

$$\Gamma_{\perp}^b = \frac{\cos \theta_i - \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}}{\cos \theta_i + \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}}$$

$$T_{\perp}^b = \frac{2 \cos \theta_i}{\cos \theta_i + \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}}$$

When  $\epsilon_2 > \epsilon_1$   $\rightarrow \Gamma_{\perp}^b$  is real & negative for all  $\theta_i$   
going from low  $\epsilon_1$  to higher  $\epsilon_2$   $\rightarrow T_{\perp}^b$  is real & positive for all  $\theta_i$   
See Figure S-3 for trends

$$\underline{\epsilon_2 = \epsilon_1} \Rightarrow \Gamma_{\perp}^b = 0 \text{ \& } T_{\perp}^b = 1 \text{ (no boundary)}$$



**Figure 5-3** Magnitude of coefficients for perpendicular polarization as a function of incident angle. (a) Reflection. (b) Transmission.

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### 5.3.1 cont.

When  $\underline{\epsilon_2 < \epsilon_1}$  or  $\frac{\epsilon_2}{\epsilon_1} < 1$ , there is a range (high to low)

of angles  $\underline{\theta_i \leq \theta_c}$  where  $\Gamma_{\perp}^b$  &  $T_{\perp}^b$  are real. However, at the critical angle  $\theta_c$ ,

$$|\Gamma_{\perp}^b| = 1 \quad \left( \begin{array}{l} \text{Total internal, to } \epsilon_1 \text{ media,} \\ \text{reflection!} \end{array} \right)$$

$\frac{\epsilon_2}{\epsilon_1} < 1$   
 +  $\theta_i = \theta_c$

For  $\theta_i > \theta_c$ ,  $\Gamma_{\perp}^b$  &  $T_{\perp}^b$  become complex.

We'll look at this more in section 5.3.4.

Lastly,  $\bar{E}_{\perp}^i = \bar{E}_{\perp}^t + \bar{E}_{\perp}^r \quad z < 0$  (media 1)

$$= \hat{a}_y E_0 e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} \left\{ 1 + \Gamma_{\perp}^b e^{j2\beta_1 z \cos \theta_i} \right\}$$

$\uparrow$   
 Traveling  
wave

$\uparrow$   
 Standing  
wave

where we define

$$\underline{\Gamma_{\perp}^b(z)} = \underline{\Gamma_{\perp}^b} e^{j2\beta_1 z \cos \theta_i} \quad z < 0$$

### 5.3.2 Parallel (Vertical or H) Polarization

→ In this case, see Fig 5-4,  $\vec{E}$  is in plane of incidence

$$\vec{H}_{11}^i = \hat{a}_y H_{11}^i e^{-j\vec{\beta}^i \cdot \vec{r}} = \hat{a}_y \frac{E_0}{\eta_1} e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} \quad z < 0 \quad (5-20b)$$

and

$$\begin{aligned} \vec{E}_{11}^i &= (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) E_0 e^{-j\vec{\beta}^i \cdot \vec{r}} \quad z < 0 \\ &= (\hat{a}_x \cos \theta_i - \hat{a}_z \sin \theta_i) E_0 e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} \quad (5-20a) \end{aligned}$$

$$\text{Where } E_{11}^i = E_0 \quad \& \quad H_{11}^i = \frac{E_{11}^i}{\eta_1} = \frac{E_0}{\eta_1}$$

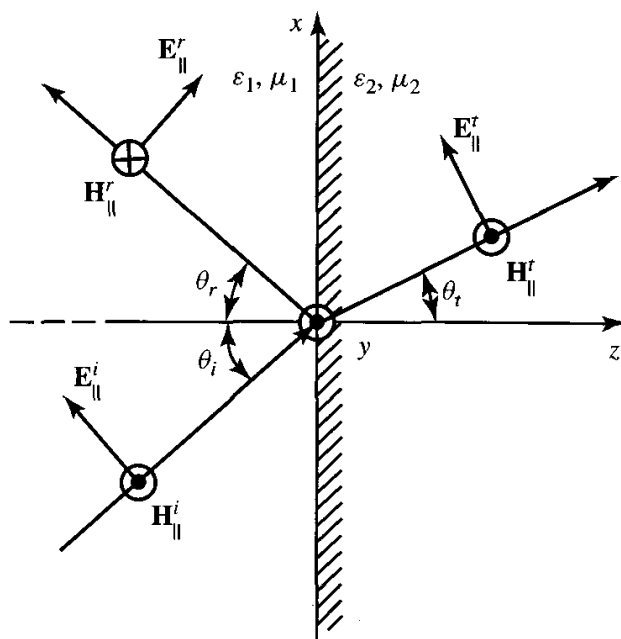
→ The reflected fields are defined

$$\begin{aligned} \vec{E}_{11}^r &= (\hat{a}_x \cos \theta_r + \hat{a}_z \sin \theta_r) E^r e^{-j\vec{\beta}^r \cdot \vec{r}} \\ &= (\hat{a}_x \cos \theta_r + \hat{a}_z \sin \theta_r) \Gamma_{11}^b E_0 e^{-j\beta_1(x \sin \theta_r - z \cos \theta_r)} \quad (5-21a) \end{aligned}$$

$$\begin{aligned} \vec{H}_{11}^r &= -\hat{a}_y H_{11}^r e^{-j\vec{\beta}^r \cdot \vec{r}} = -\hat{a}_y \frac{\Gamma_{11}^b E_0}{\eta_1} e^{-j\beta_1(x \sin \theta_r - z \cos \theta_r)} \\ & \quad z < 0 \quad (5-21b) \end{aligned}$$

$$\text{Where } E_{11}^r = \Gamma_{11}^b E^i = \Gamma_{11}^b E_0$$

$$\& \quad H_{11}^r = \frac{E_{11}^r}{\eta_1} = \frac{\Gamma_{11}^b E_0}{\eta_1}$$



**Figure 5-4** Parallel (vertical) polarized uniform plane wave incident at an oblique angle on an interface.

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5.3.2 cont.

The transmitted fields are defined

$$\vec{E}_{11}^t = (\hat{a}_x \cos \theta_t - \hat{a}_z \sin \theta_t) E_{11}^t e^{-j\beta^t \cdot \vec{r}} \quad (5-22a)$$

$$= (\hat{a}_x \cos \theta_t - \hat{a}_z \sin \theta_t) T_{11}^b E_0 e^{-j\beta_2(x \sin \theta_t + z \cos \theta_t)}$$

$$\vec{H}_{11}^t = \hat{a}_y H_{11}^t e^{j\beta^t \cdot \vec{r}} = \hat{a}_y \frac{T_{11}^b E_0}{\eta_2} e^{-j\beta_2(x \sin \theta_t + z \cos \theta_t)} \quad z > 0 \quad (5-22b)$$

where  $E_{11}^t = T_{11}^b E^i = T_{11}^b E_0$  &  $H_{11}^t = \frac{E_{11}^t}{\eta_2} = \frac{T_{11}^b E_0}{\eta_2}$

To determine  $\Gamma_{11}^b$ ,  $T_{11}^b$ ,  $\theta_r$  &  $\theta_i$ , we apply the tangential boundary conditions @  $z=0$   
 $x$  &  $y$  components are tangential

$$\vec{E}_{11t} \Big|_{z=0} = (\vec{E}_{11}^i + \vec{E}_{11}^r) \Big|_{z=0} = \vec{E}_{11}^t \Big|_{z=0} = \vec{E}_{21t} \Big|_{z=0}$$

$$\hat{a}_x \left( \cos \theta_i E_0 e^{-j\beta_1 x \sin \theta_i} + \cos \theta_r \Gamma_{11}^b E_0 e^{-j\beta_1 x \sin \theta_r} \right) = \hat{a}_x \cos \theta_t T_{11}^b E_0 e^{-j\beta_2 x \sin \theta_t}$$

$$\vec{H}_{11t} \Big|_{z=0} = (\vec{H}_{11}^i + \vec{H}_{11}^r) \Big|_{z=0} = \vec{H}_{11}^t \Big|_{z=0} = \vec{H}_{21t} \Big|_{z=0}$$

$$\hat{a}_y \left( \frac{E_0}{\eta_1} e^{-j\beta_1 x \sin \theta_i} - \frac{\Gamma_{11}^b E_0}{\eta_1} e^{-j\beta_1 x \sin \theta_r} \right) = \hat{a}_y \frac{T_{11}^b E_0}{\eta_2} e^{-j\beta_2 x \sin \theta_t}$$

S.3.2 cont.

Removing common terms, we get

$$\cos\theta_i e^{-j\beta_1 x \sin\theta_i} + \Gamma_{11}^b \cos\theta_r e^{-j\beta_1 x \sin\theta_r} = T_{11}^b \cos\theta_t e^{-j\beta_2 x \sin\theta_t} \quad (S-23a)$$

$$\frac{1}{\eta_1} \left( e^{-j\beta_1 x \sin\theta_i} - \Gamma_{11}^b e^{-j\beta_1 x \sin\theta_r} \right) = \frac{1}{\eta_2} T_{11}^b e^{-j\beta_2 x \sin\theta_t} \quad (S-23b)$$

Again, @  $z=0$ , w/  $\eta_1, \eta_2, \Gamma_{11}^b$ , &  $T_{11}^b$  being constant, (S-23b) can only be true for all  $x$  if

$$\beta_1 x \sin\theta_i = \beta_1 x \sin\theta_r = \beta_2 x \sin\theta_t$$

Snell's laws of reflection & refraction	$\Downarrow$ $\frac{\theta_i = \theta_r}{(S-24a)}$	$\Downarrow$ $\frac{\beta_1 \sin\theta_i = \beta_2 \sin\theta_t}{(S-24b)}$
---	---	---

With the now identical exponentials removed, we get

$$\cos\theta_i + \Gamma_{11}^b \cos\theta_r = T_{11}^b \cos\theta_t$$

$$\frac{1}{\eta_1} (1 - \Gamma_{11}^b) = \frac{1}{\eta_2} T_{11}^b$$

which can be solved for  $\Gamma_{11}^b$  &  $T_{11}^b$

S.3.2 cont.

$$\Gamma_{||}^b = \frac{-\eta_1 \cos \theta_i + \eta_2 \cos \theta_t}{\eta_1 \cos \theta_i + \eta_2 \cos \theta_t}$$

$$T_{||}^b = \frac{2 \eta_2 \cos \theta_i}{\eta_1 \cos \theta_i + \eta_2 \cos \theta_t}$$

Fresnel  
reflection (S-24c)  
+  
transmission (S-24d)  
coefficients  
(parallel polarization)

For dielectric media where  $\mu_1 \approx \mu_2 \approx \mu_0$

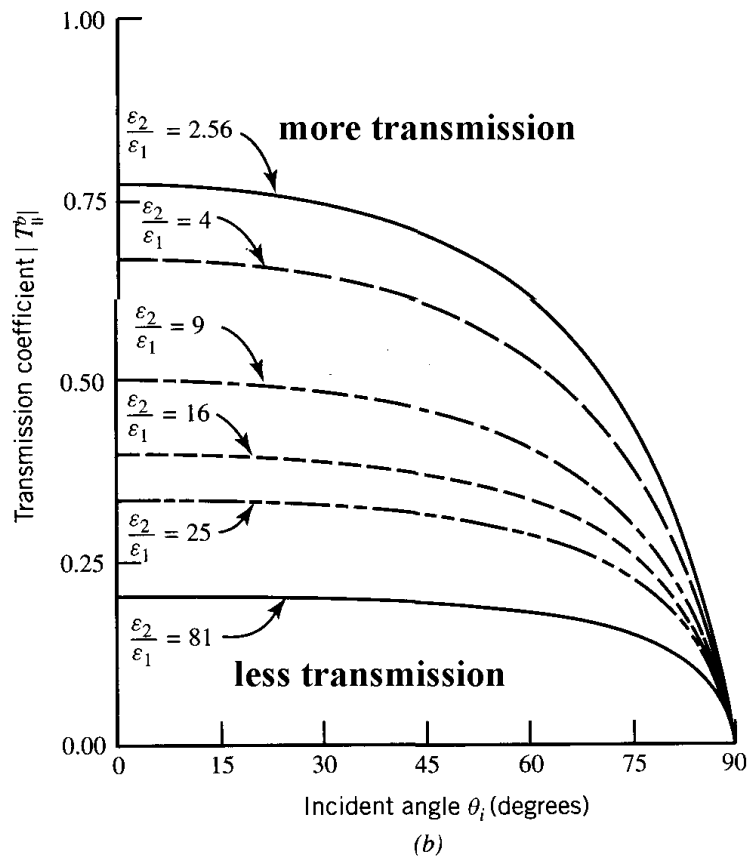
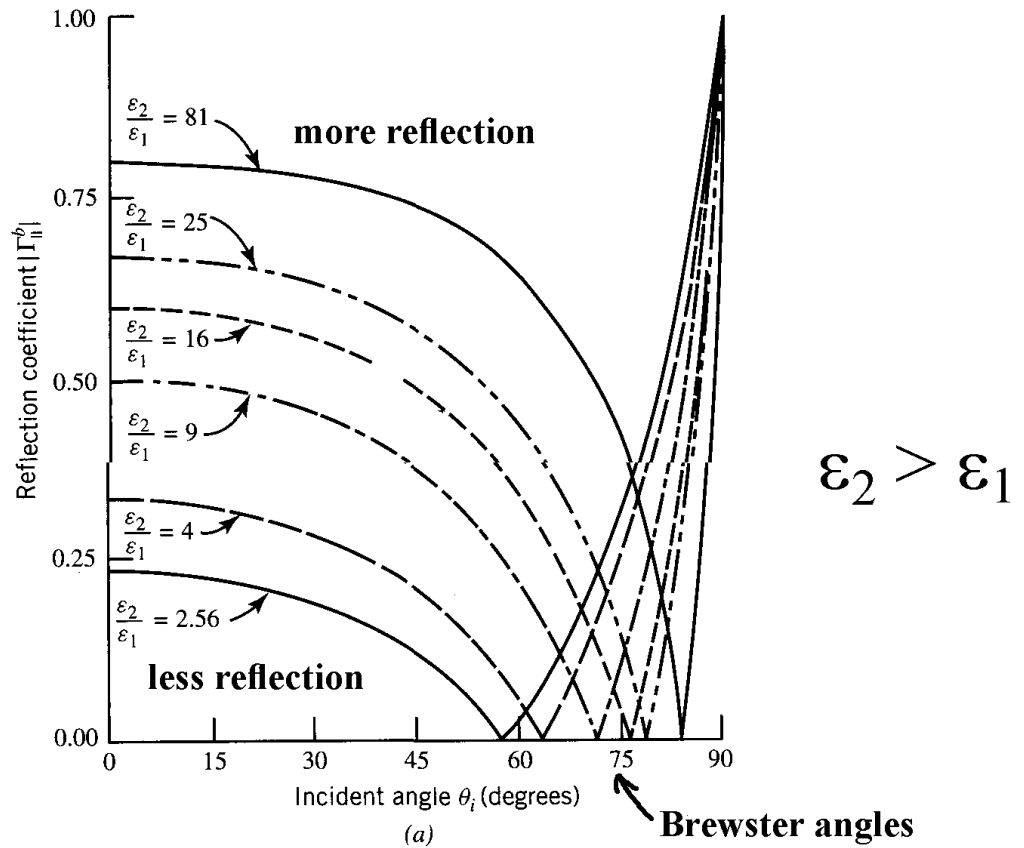
$$\left. \Gamma_{||}^b \right|_{\mu_1 = \mu_2} = \frac{-\cos \theta_i + \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}}{\cos \theta_i + \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}} \quad (S-25c)$$

$$\left. T_{||}^b \right|_{\mu_1 = \mu_2} = \frac{2 \sqrt{\frac{\epsilon_1}{\epsilon_2}} \cos \theta_i}{\cos \theta_i + \sqrt{\frac{\epsilon_1}{\epsilon_2}} \sqrt{1 - \left(\frac{\epsilon_1}{\epsilon_2}\right) \sin^2 \theta_i}} \quad (S-25d)$$

\* For  $\underline{\epsilon_2 > \epsilon_1}$ , notice in Fig 5-5a that  
(low to high)

there is an angle where  $|\Gamma_{||}^b| = 0$ ,  
i.e., the Brewster angle  $\theta_B = \theta_i$ .

\* Note that  $\theta_B$  gets larger as  $\frac{\epsilon_2}{\epsilon_1}$  increases.



**Figure 5-5** Magnitude of coefficients for parallel polarization as a function of incident angle. (a) Reflection. (b) Transmission.

*Advanced Engineering Electromagnetics* (Second Edition), Balanis, Wiley, 2012, ISBN-10: 0470589485, ISBN-13: 978-0470589489.

S.3.2 cont.

$\boxed{\epsilon_2 > \epsilon_1} \rightarrow \Gamma_{11}^b + T_{11}^b$  are both real. Further,

for  $\theta_i < \theta_B$ ,  $\Gamma_{11}^b < 0$ , while for  $\theta_i > \theta_B$ ,

$\Gamma_{11}^b > 0$ .  $T_{11}^b$  is greater than zero for all  $\theta_i$ .

$\boxed{\epsilon_2 = \epsilon_1} \Rightarrow \Gamma_{11}^b = 0$  and  $T_{11}^b = 1$  (no boundary)

$\boxed{\epsilon_2 < \epsilon_1} \rightarrow$  Again there will be a critical (high to low) angle  $\theta_c$  where  $|\Gamma_{11}^b|_{\epsilon_2 < \epsilon_1} = 1$

$\rightarrow \theta_i < \theta_c$   $\Gamma_{11}^b + T_{11}^b$  are real

$\rightarrow \theta_i > \theta_c$   $\Gamma_{11}^b + T_{11}^b$  are complex

$$\begin{aligned} \vec{E}_1^1 = \vec{E}_{11}^i + \vec{E}_{11}^r = & \hat{a}_x \cos \theta_i E_0 e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} (1 + \Gamma_{11}^b e^{j2\beta_1 z \cos \theta_i}) \\ & - \hat{a}_z \sin \theta_i E_0 e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} (1 - \Gamma_{11}^b e^{j2\beta_1 z \cos \theta_i}) \end{aligned}$$

↑  
Traveling

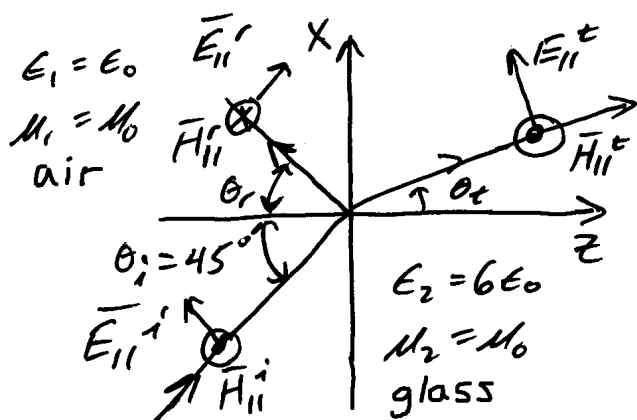
↑  
standing

$$\begin{aligned} = & \hat{a}_x \cos \theta_i E_0 e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} [1 + \Gamma_{11}(z)] \\ & - \hat{a}_z \sin \theta_i E_0 e^{-j\beta_1(x \sin \theta_i + z \cos \theta_i)} [1 - \Gamma_{11}(z)] \end{aligned}$$

where  $\Gamma_{11}(z) = \Gamma_{11}^b e^{j2\beta_1 z \cos \theta_i}$   $z < 0$



**Example-** A UPW in air ( $\epsilon_0, \mu_0, z < 0$ ) is obliquely incident on a glass half-space ( $6\epsilon_0, \mu_0, z > 0$ ) at an angle of  $45^\circ$ . The 800 MHz incident magnetic field is oriented in the y-direction and has a field strength of 0.4 mA/m at  $z = 0$ . Analyze and determine the various associated fields, power densities, and other related quantities.



$$\eta_1 = \eta_0 = \underline{376.7303 \Omega}$$

$$\beta_1 = \omega \sqrt{\mu_0 \epsilon_0} = \frac{2\pi \cdot 800 \times 10^6}{2.9979 \times 10^8} = \underline{16.76676 \frac{\text{rad}}{\text{m}}}$$

$$\eta_2 = \frac{\eta_0}{\sqrt{\epsilon_{r2}}} = \frac{376.7303}{\sqrt{6}} = \underline{153.7995 \Omega}$$

$$\beta_2 = \omega \sqrt{\mu_0 6\epsilon_0} = 2\pi (800 \times 10^6) \sqrt{4\pi \times 10^{-7} (6) 8.8541878 \times 10^{-12}}$$

$$\beta_2 = \underline{41.070007 \frac{\text{rad}}{\text{m}}}$$

Per (5-24a),  $\underline{\theta_r = \theta_i = 45^\circ}$

Per (5-24b),  $\beta_1 \sin \theta_i = \beta_2 \sin \theta_t$

$$16.76676 \sin 45^\circ = 41.07 \sin \theta_t$$

$$\hookrightarrow \theta_t = \sin^{-1} \left( \frac{16.76676 \sin 45^\circ}{41.07001} \right) = \sin^{-1}(0.288675)$$

$$= 16.778655^\circ \Rightarrow \underline{\theta_t = 16.7787^\circ}$$

Per (5-24c),  $\Gamma_{11}^b = \frac{-\eta_1 \cos \theta_i + \eta_2 \cos \theta_t}{\eta_1 \cos \theta_i + \eta_2 \cos \theta_t}$

$$\Gamma_{11}^b = \frac{-376.7303 \cos 45^\circ + 153.7995 \cos 16.7787^\circ}{376.7303 \cos 45^\circ + 153.7995 \cos 16.7787^\circ} = \frac{-124.432}{413.64}$$

$$\underline{\underline{\Gamma_{11}^b = -0.28802}}$$

$$T_{11}^b = \frac{2\eta_2 \cos \theta_i}{\eta_1 \cos \theta_i + \eta_2 \cos \theta_t} = \frac{2(153.7995) \cos 45^\circ}{413.64}$$

$$\underline{\underline{T_{11}^b = 0.52583}}$$

Per (S-20d),  $H_{11}^i = \frac{E_0}{\eta_1} \Rightarrow E_0 = 376.7303(0.4 \times 10^{-3})$   
 $E_0 = \underline{\underline{0.150692 \text{ V/m}}}$

$$(S-20b) \quad \underline{\underline{\vec{H}_{11}^i = \hat{a}_y 0.4 e^{-j16.767(0.707x + 0.707z)} \left(\frac{\text{mA}}{\text{m}}\right)}}$$

$$(S-20a) \quad \underline{\underline{\vec{E}_{11}^i = (0.707\hat{a}_x - 0.707\hat{a}_z) 0.1507 e^{-j16.767(0.707x + 0.707z)} \left(\frac{\text{V}}{\text{m}}\right)}}$$

$$(S-21a) \quad \vec{E}_{11}^r = (0.707\hat{a}_x + 0.707\hat{a}_z) 0.1507(-0.288) e^{-j16.767(0.707x - 0.707z)}$$

$$\underline{\underline{\vec{E}_{11}^r = (0.707\hat{a}_x + 0.707\hat{a}_z)(-43.402) e^{-16.767(0.707x - 0.707z)} \left(\frac{\text{mV}}{\text{m}}\right)}}$$

$$(S-21b) \quad \vec{H}_{11}^r = -\hat{a}_y \frac{-43.402 \times 10^{-3}}{376.7307} e^{-j16.767(0.707x - 0.707z)}$$

$$\underline{\underline{\vec{H}_{11}^r = +\hat{a}_y 0.1152 e^{-j16.767(0.707x - 0.707z)} \left(\frac{\text{mA}}{\text{m}}\right)}}$$

$$(S-22a) \bar{E}_{11}^t = (\hat{a}_x \cos 16.78^\circ - \hat{a}_z \sin 16.78^\circ)(0.526)(0.1507) e^{-j41.07(x \sin 16.9^\circ + z \cos 16.9^\circ)}$$

$$\bar{E}_{11}^t = (0.9574 \hat{a}_x - 0.2887 \hat{a}_z)(79.268) e^{-j41.07(0.2887x + 0.9574z)} \left( \frac{mV}{m} \right)$$


---

$$(S-22b) \bar{H}_{11}^t = \hat{a}_y \frac{0.52583(0.150692)}{153.7995} e^{-j41.07(0.2887x + 0.9574z)}$$

$$\bar{H}_{11}^t = \hat{a}_y 0.5152 e^{-41.07(0.2887x + 0.9574z)} \left( \frac{mA}{m} \right)$$


---

$$(S-26a) \Gamma_{11}(z) = \Gamma_{11}^b e^{j2\beta_1 z \cos \theta_1}$$

$$= -0.28802 e^{j2(16.76676) \cos 45^\circ z}$$

$$\Gamma_{11}(z) = -0.28802 e^{j23.7118z} \quad z \leq 0$$


---

$$(S-26) \bar{E}_{11}^i = \hat{a}_x \cos 45^\circ (0.1507) e^{-j16.767(0.707x + 0.707z)} [1 + \Gamma_{11}(z)]$$

$$- \hat{a}_z \sin 45^\circ (0.1507) e^{-j16.767(0.707x + 0.707z)} [1 - \Gamma_{11}(z)]$$

$$\bar{E}_{11}^i = \hat{a}_x 0.10656 e^{-j16.767(0.707x + 0.707z)} [1 + \Gamma_{11}(z)]$$

$$- \hat{a}_z 0.10656 e^{-j16.767(0.707x + 0.707z)} [1 - \Gamma_{11}(z)] \left( \frac{V}{m} \right)$$

$$z \leq 0$$


---

### 5.3.3 Total Transmission - Brewster Angle

Is there angle of incidence or set of conditions where the reflection coefficient goes to zero?

#### A. Perpendicular (Horizontal) Polarization

Here set  $\Gamma_{\perp}^b = \frac{\eta_2 \cos \theta_i - \eta_1 \cos \theta_t}{\eta_2 \cos \theta_i + \eta_1 \cos \theta_t} = 0$  (5-27)

algebra  $\hookrightarrow \cos \theta_i = \sqrt{\frac{\mu_1 \epsilon_2}{\mu_2 \epsilon_1}} \cos \theta_t$  (5-27a)

use  $\cos^2 A = 1 - \sin^2 A$  and square both sides

$$1 - \sin^2 \theta_i = \frac{\mu_1 \epsilon_2}{\mu_2 \epsilon_1} (1 - \sin^2 \theta_t)$$

Next use  $\beta_1 \sin \theta_i = \beta_2 \sin \theta_t \Rightarrow \sin^2 \theta_t = \frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i$

$$1 - \sin^2 \theta_i = \frac{\mu_1 \epsilon_2}{\mu_2 \epsilon_1} \left( 1 - \frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i \right)$$

algebra  $\hookrightarrow \sin \theta_i = \sqrt{\frac{\frac{\epsilon_2}{\epsilon_1} - \frac{\mu_2}{\mu_1}}{\frac{\mu_1}{\mu_2} - \frac{\mu_2}{\mu_1}}}$  (5-28a)

For most dielectrics,  $\mu_1 = \mu_2 = \mu_0$ , which implies  $\sin \theta_i \rightarrow \infty \Rightarrow$  NOT POSSIBLE

$\Rightarrow \Gamma_{\perp}^b = 0$  can't be achieved.

S.3.3 cont.B. Parallel (Vertical) Polarization

Here, set  $\Gamma_{||}^b = \frac{-\eta_1 \cos \theta_i + \eta_2 \cos \theta_t}{\eta_1 \cos \theta_i + \eta_2 \cos \theta_t} = 0$  (5-30)

algebra  $\hookrightarrow \cos \theta_i = \sqrt{\frac{\mu_2}{\mu_1} \left( \frac{\epsilon_1}{\epsilon_2} \right)} \cos \theta_t$  (5-30a)

$$\cos^2 \theta_i = 1 - \sin^2 \theta_i = \frac{\mu_2}{\mu_1} \left( \frac{\epsilon_1}{\epsilon_2} \right) \cos^2 \theta_t = \frac{\mu_2}{\mu_1} \left( \frac{\epsilon_1}{\epsilon_2} \right) (1 - \sin^2 \theta_t)$$

and use  $\sin^2 \theta_t = \frac{\mu_1}{\mu_2} \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i$

$$1 - \sin^2 \theta_i = \frac{\mu_2}{\mu_1} \frac{\epsilon_1}{\epsilon_2} \left( 1 - \frac{\mu_1}{\mu_2} \frac{\epsilon_1}{\epsilon_2} \sin^2 \theta_i \right)$$

algebra  $\hookrightarrow$

$$\sin \theta_i = \sqrt{\frac{\frac{\epsilon_2}{\epsilon_1} - \frac{\mu_2}{\mu_1}}{\frac{\epsilon_2}{\epsilon_1} - \frac{\epsilon_1}{\epsilon_2}}} \quad (5-31a)$$

$\theta_i$  real if  $\frac{\mu_2}{\mu_1} \geq \frac{\epsilon_1}{\epsilon_2}$

$\Rightarrow$  If  $\mu_2 \approx \mu_1 \approx \mu_0$ , this means  $\epsilon_1/\epsilon_2 < 1$  or  $\epsilon_1 < \epsilon_2$ ,  
i.e., the EM wave w/ parallel polarization is going  
from lower to higher permittivity media

Solve  
for  
Brewster  
angle  
 $\theta_B$

$$\theta_i = \theta_B = \sin^{-1} \sqrt{\frac{\epsilon_2}{\epsilon_1 + \epsilon_2}} \quad (5-33)$$

$$= \cos^{-1} \sqrt{\frac{\epsilon_1}{\epsilon_1 + \epsilon_2}} \quad (5-33a)$$

$$= \tan^{-1} \sqrt{\frac{\epsilon_2}{\epsilon_1}} \quad (5-33b)$$

**Example-** For a parallel polarized UPW in air ( $\epsilon_0, \mu_0, z < 0$ ) obliquely incident on a glass half-space ( $6\epsilon_0, \mu_0, z > 0$ ), find the Brewster angle. Then, determine the reflection and refraction angles as well as the reflection and transmission coefficients.

$$\text{From earlier, } \eta_1 = 376.7303 \Omega \quad \eta_2 = 153.7995 \Omega$$

$$\beta_1 = 16.76676 \text{ rad/m} \quad \beta_2 = 41.070007 \text{ rad/m}$$

$$\text{Per (5-33), } \theta_i = \theta_B = \sin^{-1} \sqrt{\frac{6\epsilon_0}{\epsilon_0 + 6\epsilon_0}} = \sin^{-1} \sqrt{6/7}$$

$$\underline{\underline{\theta_i = \theta_B = 67.792346^\circ}}$$

$$\text{Per (5-24a), } \underline{\underline{\theta_r = \theta_i = \theta_B = 67.792346^\circ}}$$

$$\text{Per (5-24b), } \theta_t = \sin^{-1} \left( \frac{16.76676 \sin 67.79^\circ}{41.070007} \right) = \sin^{-1}(0.378)$$

$$\underline{\underline{\theta_t = 22.20765^\circ}}$$

$$\text{Per (5-24c), } \Gamma_{||}^b = \frac{-376.73 \cos 67.79^\circ + 153.8 \cos 22.21^\circ}{376.73 \cos 67.79^\circ + 153.8 \cos 22.21^\circ}$$

$$\underline{\underline{\Gamma_{||}^b = 0}}$$

$$\text{Per (5-24d), } T_{||}^b = \frac{2(153.8) \cos 67.79^\circ}{376.73 \cos 67.79^\circ + 153.8 \cos 22.21^\circ}$$

$$T_{||}^b = \frac{116.261}{284.7918}$$

$$\underline{\underline{T_{||}^b = 0.40825}}$$