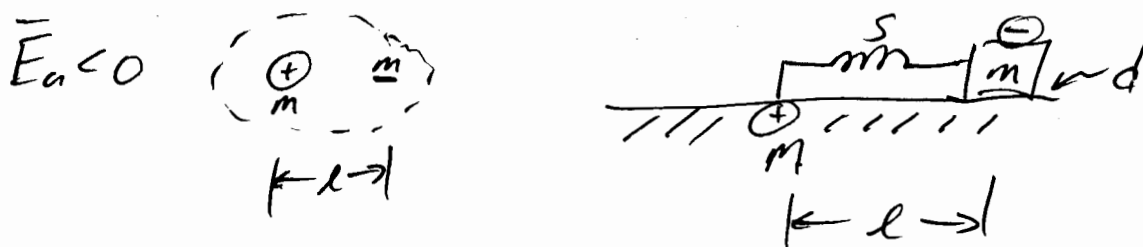
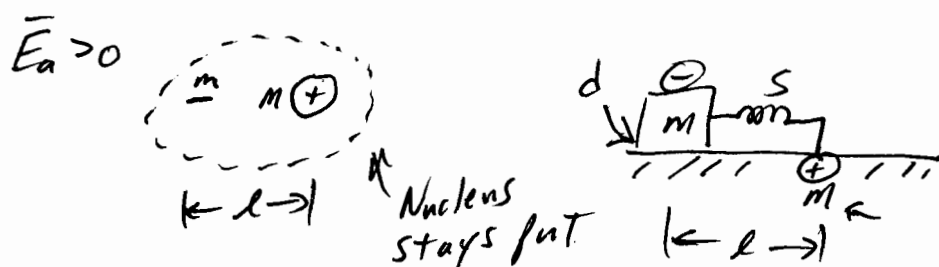


2.9 AC Variations in Materials

→ ϵ, μ, σ all vary w/ frequency

2.9.1 Complex Permittivity

→ Make analogy between polarizing an atom w/ a mechanical model of a mass, attached to a spring, being moved back-n-forth



$M \equiv$ mass of nucleus $+Q$

$m \equiv$ " " electrons $-Q$

$d \equiv$ damping/friction coefficient or takes power/energy to move charges

$S \equiv$ spring/tension coefficient or stored power/energy

$l \equiv$ displacement distance for centroids of nucleus & electrons

For a time-harmonic applied electric field,
leads to a 2nd order ODE

$$M \frac{d^2 l}{dt^2} + d \frac{dl}{dt} + Sl = Q E(t) = Q E_0 e^{j\omega t}$$

"ma"
"dv"
"sl"
"driving force"

$$\frac{d^2 l}{dt^2} + 2\alpha \frac{dl}{dt} + \omega_0^2 l = \frac{Q}{m} E_0 e^{j\omega t}$$

Where $\alpha = \frac{d}{2m} + \omega_0 = \sqrt{\frac{S}{m}}$ ↑ very similar to RLC circuits

If $\alpha > \omega_0 \rightarrow$ overdamped

$\alpha = \omega_0 \rightarrow$ critically damped

$\alpha < \omega_0 \rightarrow$ underdamped

complementary (source-free) Transient \downarrow Two solns \downarrow Particular soln

$l_c(t) = \begin{cases} \text{underdamped} \\ \text{critically damped} \\ \text{overdamped} \end{cases}$ + $l_p(t) = l_0 e^{j\omega t}$

where $l_0 = \frac{Q/m E_0}{(\omega_0^2 - \omega^2) + j\omega(d/m)}$

$$l(t) = l_c(t) + l_p(t)$$

↑ goes to zero as $t \rightarrow \infty$ as long as $d \neq 0$

Assuming an underdamped system ($\alpha < \omega_0$),
we can determine the damped resonant frequency (natural frequency) to be

$$\omega_d = \sqrt{\omega_0^2 - \alpha^2} = \sqrt{\frac{S}{m} - \left(\frac{d}{2m}\right)^2}$$

2.9.1 cont.

If $d=0$ (no damping), $\omega_d = \omega_0 = \sqrt{S/m}$.

At steady-state (i.e., $l_c(t) \rightarrow 0$), the time-varying polarization is

$$P = \mathcal{P}(t) = N_e Q l(t) = \frac{N_e \left(\frac{Q^2}{m}\right) \overbrace{E_0}^{E(t)} e^{j\omega t}}{(\omega_0^2 - \omega^2) + j\omega(d/m)}$$

↑
dipoles per unit volume

Back in section 2.2, $\bar{D} = \epsilon_0 \bar{E}_a + \bar{P} = \epsilon_0 \bar{E}_a + \epsilon_0 \chi_e \bar{E}_a$

Now $\frac{P}{E} = \frac{N_e \left(\frac{Q^2}{m}\right)}{(\omega_0^2 - \omega^2) + j\omega(d/m)}$ \leftarrow complex #!

$d E = E_0 + \frac{|P|}{|E_a|}$

and complex permittivity $\hat{\epsilon} = \epsilon_0 + \frac{P}{E} = \epsilon_0 + \frac{N_e \left(\frac{Q^2}{m}\right)}{(\omega_0^2 - \omega^2) + j\omega(d/m)}$

$= \epsilon' - j\epsilon''$

relative complex permittivity $\hat{\epsilon}_r = \frac{\hat{\epsilon}}{\epsilon_0} = \epsilon_r' - j\epsilon_r'' = 1 + \frac{\frac{N_e Q^2}{\epsilon_0 m}}{(\omega_0^2 - \omega^2) + j\omega(d/m)}$

w/ $\epsilon_r' = 1 + \frac{\frac{N_e Q^2}{\epsilon_0 m} (\omega_0^2 - \omega^2)}{(\omega_0^2 - \omega^2)^2 + (\omega d/m)^2}$

$\epsilon_r'' = \frac{N_e Q^2}{\epsilon_0 m} \left[\frac{\omega d/m}{(\omega_0^2 - \omega^2)^2 + (\omega d/m)^2} \right]$

↑ multiply top & bottom by complex conj. of denom.

2.9.1 cont.

define complex index of refraction $\equiv \dot{n} = \sqrt{\dot{\epsilon}_r}$

Note: If $d=0$, $\epsilon_r' = 1 + \frac{N_e Q^2}{\epsilon_{om}(\omega_0^2 - \omega^2)}$ & $\epsilon_r'' = 0$

How does having $\dot{\epsilon} = \epsilon' - j\epsilon''$ impact Maxwell's Equations? Ampere's Law now becomes (phasor form)

$$\begin{aligned}\vec{\nabla} \times \vec{H} &= \vec{J}_i + \vec{J}_c + j\omega \dot{\epsilon} \vec{E} = \vec{J}_t \\ &= \vec{J}_i + \sigma_s \vec{E} + j\omega(\epsilon' - j\epsilon'') \vec{E} \\ &= \vec{J}_i + (\sigma_s + \omega\epsilon'') \vec{E} + j\omega\epsilon' \vec{E} \\ &= \vec{J}_i + \underbrace{\sigma_e}_{\text{eff. cond. current dens.}} \vec{E} + \underbrace{j\omega\epsilon'}_{\text{effective displacement current density}} \vec{E}\end{aligned}$$

Equivalent conductivity $\equiv \sigma_e = \sigma_s + \omega\epsilon'' = \sigma_s + \sigma_a$

alternating field conductivity $\equiv \sigma_a = \omega\epsilon''$

static conductivity $\equiv \sigma_s$

Some materials have modest σ_s and larger σ_a (think microwave oven)

We can express the total current density

$$\begin{aligned}\bar{J}_t &= \bar{J}_i + \sigma_e \bar{E} + j\omega \epsilon' \bar{E} = \bar{J}_i + j\omega \epsilon' \left(1 - j \frac{\sigma_e}{\omega \epsilon'}\right) \bar{E} \\ &= \bar{J}_i + j\omega \epsilon' (1 - j \tan \delta_e) \bar{E}\end{aligned}$$

Define

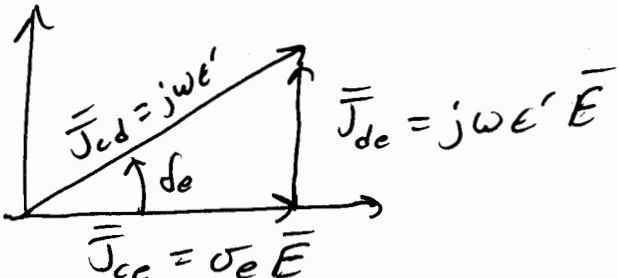
$$\begin{aligned}\text{effective loss tangent} &\equiv \tan \delta_e = \frac{\sigma_e}{\omega \epsilon'} = \frac{\sigma_s}{\omega \epsilon'} + \frac{\omega \epsilon''}{\omega \epsilon'} \\ &= \frac{\sigma_s}{\omega \epsilon'} + \frac{\epsilon''}{\epsilon'} = \tan \delta_s + \tan \delta_a\end{aligned}$$

$$\text{static loss tangent} \equiv \tan \delta_s = \frac{\sigma_s}{\omega \epsilon'}$$

$$\text{alternating loss tangent} \equiv \tan \delta_a = \frac{\epsilon''}{\epsilon'}$$

⇒ Manufacturers of PCBs and TLs when giving conductivity σ , loss tangent $\tan \delta$, and relative permittivity ϵ_r at selected frequencies typically mean σ_e , $\tan \delta_e$, and ϵ_r' . See Table 2-5.

⇒ Look at the effective conduction + displacement current densities on complex plane

$$\begin{aligned}\bar{J}_{cd} &= \bar{J}_{ce} + \bar{J}_{de} \\ &= j\omega \epsilon' \left(1 - j \frac{\sigma_e}{\omega \epsilon'}\right) \bar{E}\end{aligned}$$


$$\Rightarrow \tan \delta_e = \frac{\sigma_e}{\omega \epsilon'}$$

TABLE 2-5 Dielectric constants and loss tangents of typical dielectric materials

Material	ϵ'_r	$\tan \delta$
Air	1.0006	
Alcohol (ethyl)	25	0.1
Aluminum oxide	8.8	6×10^{-4}
Bakelite	4.74	22×10^{-3}
Carbon dioxide	1.001	
Germanium	16	
Glass	4–7	1×10^{-3}
Ice	4.2	0.1
Mica	5.4	6×10^{-4}
Nylon	3.5	2×10^{-2}
Paper	3	8×10^{-3}
Plexiglas	3.45	4×10^{-2}
Polystyrene	2.56	5×10^{-5}
Porcelain	6	14×10^{-3}
Pyrex glass	4	6×10^{-4}
Quartz (fused)	3.8	7.5×10^{-4}
Rubber	2.5–3	2×10^{-3}
Silica (fused)	3.8	7.5×10^{-4}
Silicon	11.8	
Snow	3.3	0.5
Sodium chloride	5.9	1×10^{-4}
Soil (dry)	2.8	7×10^{-2}
Styrofoam	1.03	1×10^{-4}
Teflon	2.1	3×10^{-4}
Titanium dioxide	100	15×10^{-4}
Water (distilled)	80	4×10^{-2}
Water (sea)	81	4.64
Wood (dry)	1.5–4	1×10^{-2}

2.9.1 cont.

Ex. Rogers Corporation RO4003C hydrocarbon/ceramic laminate substrate. From the datasheet

$$\rho_s = 1.7 \times 10^{10} \Rightarrow \underline{\sigma_s = 5.88 \times 10^{-15} \text{ S/m}}$$

$$\tan \delta = \tan \delta_0 = 0.0021 \text{ @ } 2.5 \text{ GHz}$$

$$= 0.0027 \text{ @ } 10 \text{ GHz}$$

$$\epsilon_{re} = \epsilon_r' = 3.67 \text{ @ } 2.5 \text{ GHz}$$

$$= 3.651 \text{ @ } 10 \text{ GHz}$$

Find the effective conductivity, imaginary part of ϵ , and ϵ_r , the complex permittivity and relative permittivity.

@ 2.5 GHz

$$\sigma_e = \omega \epsilon' \tan \delta_0 = 2\pi (2.5 \times 10^9) 3.67 \epsilon_0 (0.0021)$$

$$\underline{\sigma_e = 0.001072 \text{ S/m} = 1.072 \text{ mS/m}}$$

$$\epsilon'' = \frac{\sigma_e - \sigma_s}{\omega} = \frac{0.001072 - 5.88 \times 10^{-15}}{2\pi (2.5 \times 10^9)} = \underline{6.824 \times 10^{-14} \text{ F/m}}$$

$$\underline{\epsilon_r'' = \frac{\epsilon''}{\epsilon_0} = 0.007707} \text{ (pretty small compared to } \epsilon_r' = 3.67)$$

@ 10 GHz

$$\sigma_e = 2\pi (10 \times 10^9) 3.651 (8.8542 \times 10^{-12}) 0.0027$$

$$\underline{\sigma_e = 0.005484 \text{ S/m} = 5.484 \text{ mS/m}}$$

$$\epsilon'' = \frac{\sigma_e - \sigma_s}{\omega} = \frac{0.005484 - 5.88 \times 10^{-15}}{2\pi (10 \times 10^9)} = \underline{8.7282 \times 10^{-14} \text{ F/m}}$$

$$\epsilon_r'' = \frac{\epsilon''}{\epsilon_0} = 0.009858 \text{ (small compared to } \epsilon_r' = 3.651)$$

Consider the extremes

1. Good dielectrics ($\tan \delta_e = \frac{\sigma_e}{\omega \epsilon'} \ll 1$)

$$\bar{J}_{cd} \approx j\omega \epsilon' \bar{E} \quad (\text{nearly all displacement current density})$$

2. Good conductors ($\tan \delta_e = \frac{\sigma_e}{\omega \epsilon'} \gg 1$)

$$\bar{J}_{cd} \approx \sigma_e \bar{E} \quad (\text{nearly all conduction current density})$$

Going back to section 2.2, it was mentioned that electric polarization could occur due to different mechanisms: 1) dipole/orientational, 2) ionic/molecular, & 3) electronic. Depending on the material, 1, 2, or 3 of these mechanisms may apply and typically would occur @ different frequencies/frequency ranges.

Writing the phasor polarization vector as a function of frequency

$$\bar{P}(\omega) = \epsilon_0 \chi_e(\omega) \bar{E}_a(\omega)$$

we'll note that the electric susceptibility is a function of frequency. Further, $\chi_e(\omega)$ must be complex since \bar{E} is complex.

$$\chi_e(\omega) = \chi_e'(\omega) - j \chi_e''(\omega)$$

$$= \left[\underset{\substack{\uparrow \\ \text{dipole}}}{\chi_{ed}'(\omega)} + \underset{\substack{\uparrow \\ \text{ionic}}}{\chi_{ei}'(\omega)} + \underset{\substack{\uparrow \\ \text{electronic}}}{\chi_{ee}'(\omega)} \right] - j \left[\underset{\substack{\uparrow \\ \text{dipole}}}{\chi_{ed}''(\omega)} + \underset{\substack{\uparrow \\ \text{ionic}}}{\chi_{ei}''(\omega)} + \underset{\substack{\uparrow \\ \text{electronic}}}{\chi_{ee}''(\omega)} \right]$$

In general, $\chi_e'(\omega) = \chi_e'(-\omega)$ real part is even

$\chi_e''(-\omega) = -\chi_e''(\omega)$ imag. part is odd

Figures 2-18 & 2-19 show general samples as to how χ_e & ϵ_r can vary with frequency

→ Note, both χ_e'' & ϵ_r'' peak @/near resonant frequencies while χ_e' & ϵ_r' dip (negative slope), called anomalous dispersion.

→ Other than near resonant frequencies $|\epsilon_r|$ & ϵ_r' exhibit positive slopes, i.e., normal dispersion.

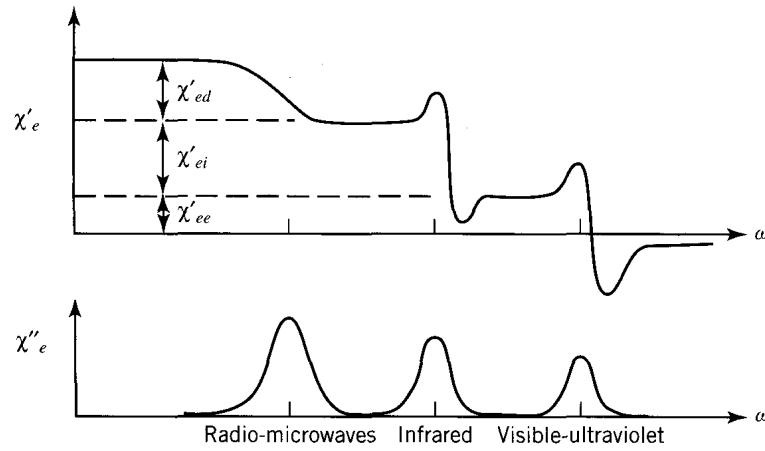


Figure 2-18 Electric susceptibility (real and imaginary) variations as a function of frequency for a typical dielectric.

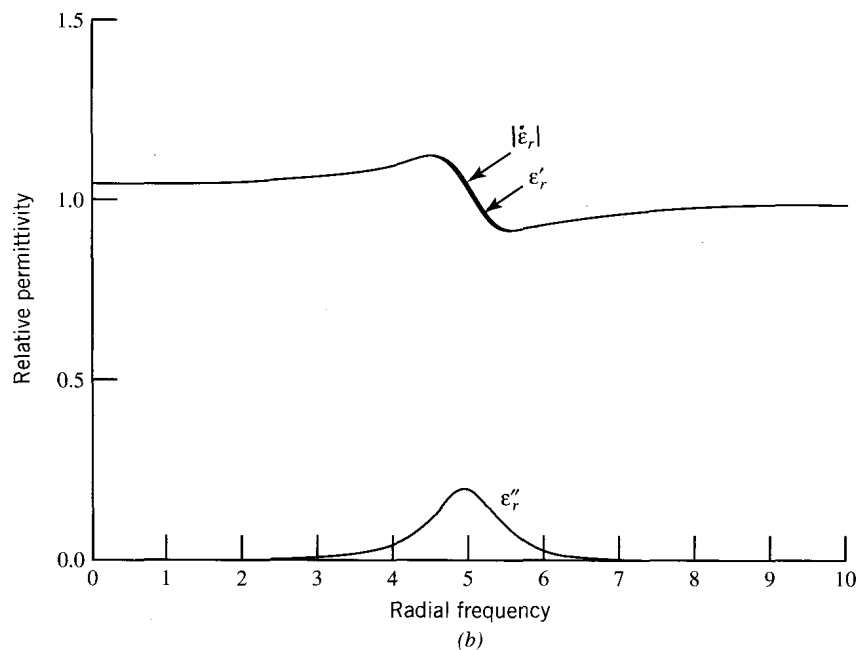
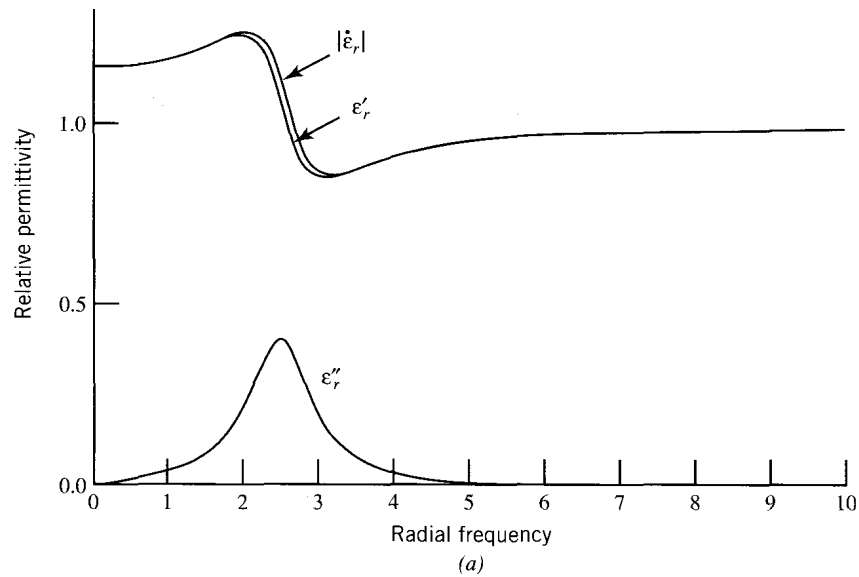


Figure 2-19 Typical frequency variations of real and imaginary parts of relative permittivity of dielectrics. (a) $N_e Q^2 / \epsilon_0 m = 1, d/m = 1, \alpha/\omega_0 = 1/5, \omega_0 = 2.5$. (b) $N_e Q^2 / \epsilon_0 m = 1, d/m = 1, \alpha/\omega_0 = 1/10, \omega_0 = 5$.

For our mechanical analogy, we assumed a single resonant frequency. However, actual material can have multiple resonances due to different mechanisms. To allow for this possibility, or P resonances, let-

$$\frac{P}{\epsilon} = \sum_{s=1}^P \frac{N_e \frac{Q^2}{m}}{(\omega_s^2 - \omega^2) + j \frac{\omega d}{m}}$$

$$\dot{\epsilon} = \epsilon' - j\epsilon'' = \epsilon_0 + \sum_{s=1}^P \frac{N_e \frac{Q^2}{m}}{(\omega_s^2 - \omega^2) + j \frac{\omega d}{m}}$$

$$\dot{\epsilon}_r = \epsilon_r' - j\epsilon_r'' = 1 + \sum_{s=1}^P \frac{N_e \frac{Q^2}{\epsilon_0 m}}{(\omega_s^2 - \omega^2) + j \frac{\omega d}{m}}$$

$$\epsilon_r' = 1 + \sum_{s=1}^P \frac{N_e Q^2}{\epsilon_0 m} \frac{(\omega_s^2 - \omega^2)}{(\omega_s^2 - \omega^2)^2 + (\omega d/m)^2}$$

$$\epsilon_r'' = \sum_{s=1}^P \frac{N_e Q^2}{\epsilon_0 m} \frac{\omega d/m}{(\omega_s^2 - \omega^2)^2 + (\omega d/m)^2}$$

$\Rightarrow \epsilon_r'$ & ϵ_r'' are related by the Kramers-Kronig equations

$$\epsilon_r'(\omega) = 1 + \frac{2}{\pi} \int_0^{\infty} \frac{\omega' \epsilon_r''(\omega')}{\omega'^2 - \omega^2} d\omega'$$

$$\epsilon_r''(\omega) = \frac{2\omega}{\pi} \int_0^{\infty} \frac{1 - \epsilon_r'(\omega')}{\omega'^2 - \omega^2} d\omega'$$

For many materials, the Debye equation can be used to find the complex relative permittivity (as well as $\sigma_e(\omega)$)

$$\dot{\epsilon}_r(\omega) = \epsilon_{r\infty}' + \frac{\epsilon_{rs}' - \epsilon_{r\infty}'}{1 + j\omega\tau_e}$$

where

new relaxation time constant $\equiv \tau_e = \tau \frac{\epsilon_{rs}' + 2}{\epsilon_{r\infty}' + 2}$

$$\epsilon_r'(\omega) = \epsilon_{r\infty}' + \frac{\epsilon_{rs}' - \epsilon_{r\infty}'}{1 + (\omega\tau_e)^2}$$

$$\Rightarrow \omega_e = \frac{1}{\tau_e}$$

$$f_e = \frac{1}{2\pi\tau_e}$$

$$\epsilon_r''(\omega) = \frac{(\epsilon_{rs}' - \epsilon_{r\infty}')\omega\tau_e}{1 + (\omega\tau_e)^2}$$

relaxation frequencies

provided that the following are known

$\epsilon_{rs}' \equiv$ static relative permittivity ($\omega=0$)

$\epsilon_{r\infty}' \equiv$ relative permittivity as $\omega \rightarrow \infty$

τ_e or τ

$\sigma_0 \equiv$ static conductivity (S/m)

\Rightarrow This is good for a single resonance.

More terms can be added to account for multiple resonances.