

**8.6** Solve the design equations in Section 8.3 for the elements of an  $N = 2$  equal-ripple filter if the ripple specification is 1.0 dB.

- 8.6 with ripple specification changed to 1.5 dB. a) Find the parameter  $k^2$  as well as values  $k_{\text{neg}}$  &  $k_{\text{pos}}$ . b) Find the equations for the  $P_{LR}$  in terms of  $\omega$  and the circuit component values. c) Find the two possible values for  $R$ , i.e.,  $R_{\text{small}}$  &  $R_{\text{big}}$ . d) Find equations for realizable values of the inductance  $L_{k\text{neg}}$  and  $L_{k\text{pos}}$  in terms of  $R$ ,  $C$ , and  $k_{\text{neg}}$  or  $k_{\text{pos}}$ . e) Equate  $\omega^2$  coefficients and then use the  $L_{k\text{neg}}$  &  $L_{k\text{pos}}$  equations to generate two polynomials in terms of  $C$ . f) Find  $R_s$ ,  $L$ ,  $C$ , &  $R$  and draw prototype low-pass filter circuit with component values. [Hints: Try all four permutations of  $k_{\text{neg}}$ ,  $k_{\text{pos}}$ ,  $R_{\text{small}}$ , and  $R_{\text{big}}$  in the polynomials to find the solution with realizable components. Consider using quadratic formula or numerical polynomial root solver.]

a) Per (8.54),  $P_{LR} = 1 + k^2 T_N^2\left(\frac{\omega}{\omega_c}\right)$  for a Chebyshev low-pass filter.

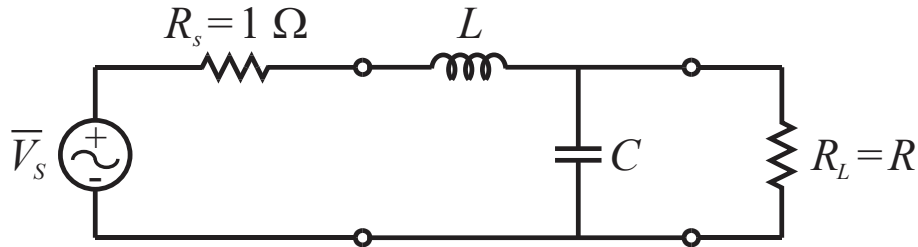
Chebyshev polynomials are equal to 1 when their argument is 1, i.e.,  $T_N^2(1) = 1$  at  $\omega = \omega_c$ .

Therefore,  $P_{LR}(\omega_c) = 1 + k^2 T_N^2\left(\frac{\omega_c}{\omega_c}\right) = 1 + k^2$ . Here, we desire

$$P_{LR}(\omega_c) = 1 + k^2 = 1.5 \text{ dB} = 10^{1.5/10} = 1.412537545 \Rightarrow \underline{k^2 = 0.412537545}.$$

Solving we get  $k = \pm 0.642290857 \Rightarrow \underline{k_{\text{pos}} = 0.642290857}$  &  $\underline{k_{\text{neg}} = -0.642290857}$ .

b) For our prototype second order low-pass filter, we let  $R_s = 1$  and assume  $\omega_c = 1$  rad/s.



Now (8.54) becomes (8.61)  $P_{LR} = 1 + k^2 T_N^2(\omega)$ . For  $N = 2$ , per (5.56b)  $T_2(x) = 2x^2 - 1$ , so  $T_2^2(\omega) = (2\omega^2 - 1)^2$ . Equating the  $P_{LR}$  equation (8.60) with (8.61) yields (8.62)

$$\begin{aligned} P_{LR} &= 1 + k^2 (2\omega^2 - 1)^2 = 1 + k^2 (4\omega^4 - 4\omega^2 + 1) \\ &= 1 + \frac{1}{4R} [(1-R)^2 + (R^2 C^2 + L^2 - 2LCR^2)\omega^2 + L^2 C^2 R^2 \omega^4] \end{aligned}$$

c) Letting  $\omega = 0$ , we get  $1 + k^2 = 1 + \frac{1}{4R} [(1-R)^2] \Rightarrow k^2 = \frac{(1-R)^2}{4R}$ .

After some algebra & using the quadratic formula, we get (8.63)  $R = 1 + 2k^2 \pm 2k\sqrt{1+k^2}$ .

For this case,  $R = 1 + 2(0.412537545) \pm 2(\pm 0.642290857)\sqrt{1.412537545}$  which has two possible solutions  $\Rightarrow \underline{R_{\text{big}} = 3.351803}$  &  $\underline{R_{\text{small}} = 0.298347}$ .

d) Equating the  $\omega^4$  coefficients of (8.62) gives  $4k^2 = \frac{L^2 C^2 R^2}{4R} \Rightarrow L^2 = \frac{4^2 k^2 R}{C^2 R^2} = \frac{16k^2}{C^2 R}$ .

Solving for the inductance,  $L = \frac{\pm 4k}{C\sqrt{R}}$  which implies  $\underline{L_{\text{kneg}} = \frac{-4k_{\text{neg}}}{C\sqrt{R}}}$  or  $\underline{L_{\text{kpos}} = \frac{4k_{\text{pos}}}{C\sqrt{R}}}$  are the realizable solutions where  $L > 0$ .

e) Equating the  $\omega^2$  coefficients of (8.62) gives

$$-4k^2 = \frac{1}{4R}(R^2 C^2 + L^2 - 2LCR^2)$$

$$-16k^2 R = R^2 C^2 + \frac{16k^2}{C^2 R} - 2LCR^2$$

$$16k^2 RC^2 + R^2 C^4 + \frac{16k^2}{R} - 2LR^2 C^3 = 0$$

If  $L_{\text{kneg}}$  is substituted in, we get a polynomial in terms of  $C$ .

$$16k^2 RC^2 + R^2 C^4 + \frac{16k^2}{R} - 2\left(\frac{-4k_{\text{neg}}}{C\sqrt{R}}\right)R^2 C^3 = 0$$

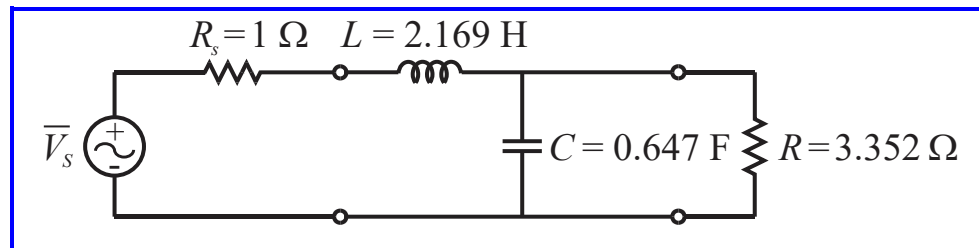
$$R^2 C^4 + \left(16k^2 R + \frac{8k_{\text{neg}} R^2}{\sqrt{R}}\right) C^2 + \frac{16k^2}{R} = 0.$$

If  $L_{\text{kpos}}$  is substituted in, we get another polynomial in terms of  $C$ .

$$16k^2 RC^2 + R^2 C^4 + \frac{16k^2}{R} - 2\left(\frac{4k_{\text{pos}}}{C\sqrt{R}}\right)R^2 C^3 = 0$$

$$R^2 C^4 + \left(16k^2 R - \frac{8k_{\text{pos}} R^2}{\sqrt{R}}\right) C^2 + \frac{16k^2}{R} = 0.$$

f) MathCAD was used to solve these polynomials for  $C$  to determine which combination(s) of  $R_{\text{big}}$ ,  $R_{\text{small}}$ ,  $k_{\text{pos}}$ , and  $k_{\text{neg}}$  lead to realizable values of  $L$  &  $C$ . As shown, only the  $R_{\text{big}}$  &  $k_{\text{new}}$  or  $R_{\text{big}}$  &  $k_{\text{pos}}$  combinations work and yield the same solutions. The 2<sup>nd</sup>-order prototype Chebyshev low-pass filter w/ a ripple of 1.5 dB is shown below. In terms of immittances:  $\underline{g_0 = 1, g_1 = 2.1688, g_2 = 0.6470, \text{ and } g_3 = 3.3518}$ .



**Case 1**  $k_{neg}$  and  $R_{big}$ -  $C$  &  $L$  are real & positive!

$$v_{case1} := \begin{pmatrix} R_{big}^2 \\ 0 \\ 16 \cdot k_{sqrd} \cdot R_{big} + \frac{8 \cdot k_{neg} \cdot R_{big}^2}{\sqrt{R_{big}}} \\ 0 \\ \frac{16 \cdot k_{sqrd}}{R_{big}} \end{pmatrix} \quad v_{case1} = \begin{pmatrix} 11.2346 \\ 0 \\ -9.4072 \\ 0 \\ 1.9693 \end{pmatrix}$$

$$f(C) := R_{big}^2 \cdot C^4 + \left( 16 \cdot k_{sqrd} \cdot R_{big} + \frac{8 \cdot k_{neg} \cdot R_{big}^2}{\sqrt{R_{big}}} \right) \cdot C^2 + \frac{16 \cdot k_{sqrd}}{R_{big}}$$

$$C := 0.5 \quad C1 := \text{root}(f(C), C) \quad L1 := \frac{-4 \cdot k_{neg}}{C1 \cdot \sqrt{R_{big}}} \quad R1 := R_{big}$$

$$R_s := 1 \quad C1 = 0.647038 \quad L1 = 2.168815 \quad R1 = 3.351803$$

**Case 2**  $k_{neg}$  and  $R_{small}$ - NO,  $C$  &  $L$  are complex

$$v_{case2} := \begin{pmatrix} R_{small}^2 \\ 0 \\ 16 \cdot k_{sqrd} \cdot R_{small} + \frac{8 \cdot k_{neg} \cdot R_{small}^2}{\sqrt{R_{small}}} \\ 0 \\ \frac{16 \cdot k_{sqrd}}{R_{small}} \end{pmatrix} \quad v_{case2} = \begin{pmatrix} 0.089 \\ 0 \\ 1.1319 \\ 0 \\ 22.1239 \end{pmatrix}$$

$$f(C) := R_{small}^2 \cdot C^4 + \left( 16 \cdot k_{sqrd} \cdot R_{small} + \frac{8 \cdot k_{neg} \cdot R_{small}^2}{\sqrt{R_{small}}} \right) \cdot C^2 + \frac{16 \cdot k_{sqrd}}{R_{small}}$$

$$C := 0.5 \quad C2 := \text{root}(f(C), C) \quad L2 := \frac{-4 \cdot k_{neg}}{C2 \cdot \sqrt{R_{small}}} \quad R2 := R_{small}$$

$$R_s = 1 \quad C2 = 2.169 - 3.326i \quad L2 = 0.647 + 0.992i \quad R2 = 0.298347$$

**Case 3**  $k_{pos}$  and  $R_{big}$ -  $C$  &  $L$  are real & positive and same values as Case 1!

$$v_{case3} := \begin{pmatrix} R_{big}^2 \\ 0 \\ 16 \cdot k_{sqrd} \cdot R_{big} - \frac{8 \cdot k_{pos} \cdot R_{big}^2}{\sqrt{R_{big}}} \\ 0 \\ \frac{16 \cdot k_{sqrd}}{R_{big}} \end{pmatrix} \quad v_{case3} = \begin{pmatrix} 11.2346 \\ 0 \\ -9.4072 \\ 0 \\ 1.9693 \end{pmatrix}$$

$$f(C) := R_{big}^2 \cdot C^4 + \left( 16 \cdot k_{sqrd} \cdot R_{big} - \frac{8 \cdot k_{pos} \cdot R_{big}^2}{\sqrt{R_{big}}} \right) \cdot C^2 + \frac{16 \cdot k_{sqrd}}{R_{big}}$$

$$C := 0.5 \quad C3 := \text{root}(f(C), C) \quad L3 := \frac{4 \cdot k_{pos}}{C3 \cdot \sqrt{R_{big}}} \quad R3 := R_{big}$$

$$R_s := 1 \quad C3 = 0.647038 \quad L3 = 2.168815 \quad R3 = 3.351803$$

**Case 4**  $k_{pos}$  and  $R_{small}$ - NO,  $C$  &  $L$  are complex

$$v_{case4} := \begin{pmatrix} R_{small}^2 \\ 0 \\ 16 \cdot k_{sqrd} \cdot R_{small} - \frac{8 \cdot k_{pos} \cdot R_{small}^2}{\sqrt{R_{small}}} \\ 0 \\ \frac{16 \cdot k_{sqrd}}{R_{small}} \end{pmatrix} \quad v_{case4} = \begin{pmatrix} 0.089 \\ 0 \\ 1.1319 \\ 0 \\ 22.1239 \end{pmatrix}$$

$$f(C) := R_{small}^2 \cdot C^4 + \left( 16 \cdot k_{sqrd} \cdot R_{small} - \frac{8 \cdot k_{pos} \cdot R_{small}^2}{\sqrt{R_{small}}} \right) \cdot C^2 + \frac{16 \cdot k_{sqrd}}{R_{small}}$$

$$C := 0.5 \quad C4 := \text{root}(f(C), C) \quad L4 := \frac{4 \cdot k_{pos}}{C4 \cdot \sqrt{R_{small}}} \quad R4 := R_{small}$$

$$R_s = 1 \quad C4 = 2.169 - 3.326i \quad L4 = 0.647 + 0.992i \quad R4 = 0.298347$$