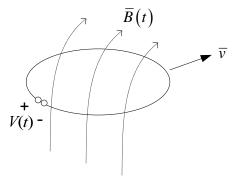
Lecture 4: Faraday's Law and Moving Circuits.

In the previous two lectures, we've been investigating Faraday's law, which is written as

$$\oint_{c(s)} \overline{E} \cdot d\overline{l} = -\frac{d}{dt} \int_{s(c)} \overline{B} \cdot d\overline{s}$$
(1)

As discussed in text Section 9.3, however, there is an alternate form of (1) that is sometimes used when the circuit moves through space. This is especially true if \overline{B} is also a function of time.

Consider the circuit shown below:



This loop is traveling with velocity \overline{v} (relative to the frame defining \overline{B}) in a time-varying B field, $\overline{B}(t)$.

The emf measured by a high-impedance voltmeter connected to this circuit is given in (1). Charges in the wire are induced to move under the influence of a force given by the Lorentz force equation

$$\overline{F} = q\left(\overline{E} + \overline{v} \times \overline{B}\right) \tag{2}$$

Motional and Transformer EMF

However, for an observer traveling with the loop, there is no apparent velocity. Instead, this observer would observe a force \overline{F}' due only to an apparent electric field \overline{E}'

$$\overline{F}' = q\overline{E}' \tag{3}$$

The two forces \overline{F} and $\overline{F'}$ must be equal, so that from (2) and (3) we find that

$$\overline{E}' = \overline{E} + \overline{\nu} \times \overline{B} \tag{4}$$

Now, let's substitute (4) back into Faraday's law (1). After considerable simplification we find that

$$\oint_{c(s)} \overline{E'} \cdot d\overline{l} = -\int_{s(c)} \frac{\partial B}{\partial t} \cdot d\overline{s} + \oint_{c(s)} \left(\overline{v} \times \overline{B} \right) \cdot d\overline{l}$$
(5)

This is an alternate form to Faraday's law (1). Notice that there is no magnetic flux in this form!

Important points to note in (5):

• $\oint_{c(s)} (\overline{v} \times \overline{B}) \cdot d\overline{l}$ is called the **motional emf**. The circuit

defined by contour c is said to "cut" the B field lines, whether or not B is constant or changing with time. (Note

that it's assumed here the shape of the circuit does not change with time.)

• $-\int_{s(c)} \frac{\partial \overline{B}}{\partial t} \cdot d\overline{s}$ is called the **transformer emf**. (Notice that

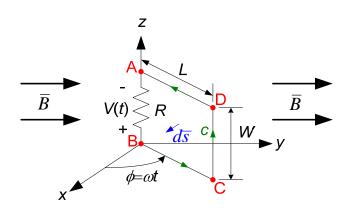
the time derivative appears inside the integral.) There is no motion present in this term.

• The same *emf* would be measured by a voltmeter whether it is moving with the circuit as in (5) or stationary as in (1) since

$$\oint_{c(s)} \frac{\overline{F}}{q} \cdot d\overline{l} = \oint_{c(s)} \frac{\overline{F'}}{q} \cdot d\overline{l} = emf$$

Either of these two equations (1) and (5) yield exactly the same *emf*. This property will be illustrated in the two examples shown next. However, one form may be preferred over another depending on the problem.

Example N4.1: Determine the induced resistor voltage V in the rotating loop circuit shown below using the form of Faraday's law given in (5). The loop is immersed in the field $\overline{B} = \hat{a}_{v}B_{o}$ [T].



The direction of c was arbitrarily chosen as shown. From (5)

$$emf = -\int_{s(c)} \frac{\partial B}{\partial t} \cdot d\overline{s} + \oint_{c(s)} \left(\overline{v} \times \overline{B} \right) \cdot d\overline{l} = \oint_{c(s)} \left(\overline{v} \times \overline{B} \right) \cdot d\overline{l}$$

The first term in the RHS is zero because \overline{B} is not a function of time.

We will consider contributions to $(\overline{v} \times \overline{B}) \cdot d\overline{l}$ from each of the four line segments that comprise the contour *c*:

- \overline{AB} : $\overline{v} = 0$ such that $(\overline{v} \times \overline{B}) \cdot d\overline{l} = 0$.
- \overline{BC} : $\overline{v} \times \overline{B} \perp d\overline{l}$ such that $(\overline{v} \times \overline{B}) \cdot d\overline{l} = 0$.
- \overline{DA} : $\overline{v} \times \overline{B} \perp d\overline{l}$ such that $(\overline{v} \times \overline{B}) \cdot d\overline{l} = 0$.

•
$$\overline{CD}$$
: $\overline{v} = \hat{a}_{\phi} L \omega$, therefore
 $\overline{v} \times \overline{B} = (\hat{a}_{\phi} L \omega) \times (\hat{a}_{y} B_{o}) = (\hat{a}_{\phi} \times \hat{a}_{y}) L \omega B_{o}$

The cross product $\hat{a}_{\phi} \times \hat{a}_{y}$ involves unit vectors from two different coordinate systems. A direct calculation approach

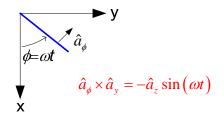
involves converting \hat{a}_{ϕ} to the Cartesian coordinate system, then performing the cross product. That is,

$$\hat{a}_{\phi} = -\hat{a}_x \sin\phi + \hat{a}_y \cos\phi \tag{6}$$

such that

$$\hat{a}_{\phi} \times \hat{a}_{y} = -\hat{a}_{x} \sin(\omega t) \times \hat{a}_{y} = -\hat{a}_{z} \sin(\omega t)$$

Another approach to this cross product calculation is more heuristic. The calculation of the cross product of unit vectors is aided by drawing a top-down sketch of the rotating loop:



Consequently,

$$\overline{v} \times \overline{B} = -\hat{a}_z L\omega B_o \sin(\omega t)$$

Therefore,

$$(\overline{v} \times \overline{B}) \cdot d\overline{l} = \left[-\hat{a}_z L \omega B_o \sin(\omega t) \right] \cdot \underbrace{\hat{a}_z dz}_{=d\overline{l}}$$
$$= -L \omega B_o \sin(\omega t) dz$$

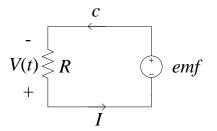
Continuing with the calculation of emf, then,

$$emf = \oint_{c(s)} \left(\overline{v} \times \overline{B}\right) \cdot d\overline{l} = -\int_{0}^{W} L\omega B_{o} \sin(\omega t) dz$$

such that

$$emf = -\omega LWB_o \sin(\omega t) [V]$$
(7)

Next, draw the equivalent lumped-element circuit:



By KVL:

 $V + emf = 0 \implies V = -emf = \omega LWB_o \sin(\omega t)$ [V]

We can interpret the results of the line integral of $(\overline{v} \times \overline{B}) \cdot d\overline{l}$ for each of the four segments in the context of which segments actually "cut" the \overline{B} field lines. Segments \overline{BC} and \overline{DA} are in motion, but since these segments do not "cut" the \overline{B} field lines, no *emf* is generated by their motion through \overline{B} . Their motion is always parallel to the \overline{B} field lines.

Conversely, the vertical section \overline{CD} is "cutting" the \overline{B} field lines as it rotates. Consequently, an *emf* is generated in this segment as it rotates through \overline{B} . Note that for $\phi = 0$ and $\phi =$ 180°, this segment does not cut the \overline{B} field lines so no *emf* is generated near these angles. **Example N4.2**: Repeat the previous example, but use the form of Faraday's law in (1).

From (1): $emf = -\frac{d\psi_m}{dt} = -\frac{d}{dt} \int_s \overline{B} \cdot d\overline{s}$

As indicated in the figure on page 3, $d\overline{s} = -\hat{a}_{\phi}drdz$. Consequently,

$$emf = +\frac{d}{dt} \int_{0}^{W} \int_{0}^{L} \left(\hat{a}_{y} \cdot \hat{a}_{\phi}\right) B_{o} dr dz$$

Similar to the previous example problem, this inner product of unit vectors can be evaluated using (6), or the heuristic approach by making a quick sketch:

Therefore,

$$emf = WLB_o \frac{d}{dt} \cos(\omega t) = -WLB_o \omega \sin(\omega t)$$
[V]

which is the same answer as in (7) of the previous example, as expected.

As shown in the previous example

$$V = -emf$$

Therefore

$$\mathbf{y}$$

$$\mathbf{a}_{\phi}$$

$$\hat{a}_{y} \cdot \hat{a}_{\phi} = \cos(\omega t)$$

$$\mathbf{x}$$

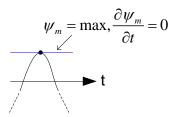
$V = -emf = WLB_o \omega \sin(\omega t)$ [V]

Once again, this is the same answer as in the previous example, as expected.

Note that here in Example N4.2 we have solved for V without resorting to motional *emf* and/or transformer *emf*, in contrast to the previous Example N4.1 (where only motional *emf* was present). Instead, we have just applied Faraday's law (1), which is always valid.

Lastly, notice that when the loop is in the *xz* plane, ψ_m is maximum, but V = 0! Further, when the loop lies in the *yz* plane, $\psi_m = 0$ but |V| is maximum! Very strange, what is happening?

Recall that the *emf* is not equal to ψ_m in Faraday's law. Rather, it is equal to the negative time-rate-of-change of ψ_m . When the loop is in the *xz* plane:



while when the loop is in the *yz* plane:

