

EE 381 Electric & Magnetic Fields Examination #1 (Fall 2025)Name KEY A

Instructions: Closed book. Put answers in indicated spaces, use notation as given in class for coordinates & vectors, & show all work for credit. Insert equation sheets in exam. $\epsilon_0 = 8.8542 \times 10^{-12} \text{ F/m}$ & $\mu_0 = 4\pi \times 10^{-7} \text{ H/m}$

- 1) A coaxial transmission line (TL) is made with conductors (ϵ_0 , μ_0 , and $\sigma_c = 8 \times 10^6 \text{ S/m}$) with the inner conductor having a diameter of 3 mm and the shield having an inner diameter of 10 mm and thickness of 0.5 mm separated by plastic insulation ($2.1\epsilon_0$, μ_0 , and $\sigma = 6 \times 10^{-6} \text{ S/m}$). Determine the skin depth and per-unit-length parameters R , L , C , and G at 9 GHz. Then, determine the characteristic impedance and propagation constant of the TL. Put complex quantities in rectangular format.

Per Table 11.1,

$$\delta_s = \frac{1}{\sqrt{\pi f \mu_c \sigma_c}} = \frac{1}{\sqrt{\pi(9 \times 10^9)4\pi \times 10^{-7}(8 \times 10^6)}} \Rightarrow \delta_s = \underline{1.87566 \times 10^{-6} \text{ m} = 1.876 \mu\text{m}}.$$

Note: The lead shield thickness $0.5 \text{ mm} \gg \delta_s$. Also, $2a = 3 \text{ mm}$ and $2b = 10 \text{ mm}$.

Using equations from Table 11.1,

$$R = \frac{1}{2\pi \delta_s \sigma_c} \left(\frac{1}{a} + \frac{1}{b} \right) = \frac{1}{2\pi(1.87566 \times 10^{-6})(8 \times 10^6)} \left(\frac{1}{0.0015} + \frac{1}{0.005} \right) \Rightarrow \underline{R = 9.19239 \Omega/\text{m}}$$

$$L = \frac{\mu}{2\pi} \ln(b/a) = \frac{4\pi \times 10^{-7}}{2\pi} \ln(10/3) \Rightarrow \underline{L = 2.40795 \times 10^{-7} \text{ H/m} = 240.795 \text{ nH/m}}$$

$$G = \frac{2\pi \sigma}{\ln(b/a)} = \frac{2\pi(6 \times 10^{-6})}{\ln(10/3)} \Rightarrow \underline{G = 3.131226 \times 10^{-5} \text{ S/m} = 31.312 \mu\text{S/m}}$$

$$C = \frac{2\pi \epsilon}{\ln(b/a)} = \frac{2\pi(2.1)8.8541878 \times 10^{-12}}{\ln(10/3)} \Rightarrow \underline{C = 9.70356 \times 10^{-11} \text{ F/m} = 97.036 \text{ pF/m}}$$

Per (11.11),

$$\gamma = \sqrt{(R + j\omega L)(G + j\omega C)} = \sqrt{(9.1924 + j2\pi(9 \times 10^9)2.4 \times 10^{-7})(3.1 \times 10^{-5} + j2\pi(9 \times 10^9)9.7 \times 10^{-11})} \Rightarrow \underline{\gamma = 0.09305 + j273.3451 \text{ 1/m}}.$$

Per (11.19), $Z_C = \sqrt{\frac{R + j\omega L}{G + j\omega C}} = \sqrt{\frac{9.1924 + j2\pi(9 \times 10^9)2.4 \times 10^{-7}}{3.1 \times 10^{-5} + j2\pi(9 \times 10^9)9.7 \times 10^{-11}}} \Rightarrow \underline{Z_0 = 49.81473 - j0.01667 \Omega}.$

$$\delta_s = \underline{1.876 \mu\text{m}} \quad R = \underline{9.1924 \Omega/\text{m}} \quad L = \underline{240.795 \text{ nH/m}} \quad C = \underline{97.036 \text{ pF/m}}$$

$$G = \underline{31.312 \mu\text{S/m}} \quad Z_C = \underline{49.81473 - j0.01667 \Omega} \quad \gamma = \underline{0.09305 + j273.3451 \text{ 1/m}}$$

2) Given vectors $\vec{H} = -3\hat{a}_x + 5\hat{a}_y - 7\hat{a}_z$ and $\vec{I} = 6\hat{a}_r$ and the spherical point $P(5, 0.36\pi, 1.7\pi)$, find:

a) the point P in Cartesian coordinates,

$$\text{Per (2.22), } x = r \sin(\theta) \cos(\phi) = 5 \sin(0.36\pi) \cos(1.7\pi) = 2.6592$$

$$y = r \sin(\theta) \sin(\phi) = 5 \sin(0.36\pi) \sin(1.7\pi) = -3.6601$$

$$z = r \cos(\theta) = 5 \cos(0.36\pi) = 2.1289$$

$$P_{\text{Cartesian}} = \underline{(2.6592, -3.6601, 2.1289)}$$

b) vector $\vec{I}$ in Cartesian coordinates at point P ,

Per (2.24), use $\hat{a}_r = \sin \theta \cos \phi \hat{a}_x + \sin \theta \sin \phi \hat{a}_y + \cos \theta \hat{a}_z$ in

$$\vec{I} = 6\hat{a}_r = 6(\sin \theta \cos \phi \hat{a}_x + \sin \theta \sin \phi \hat{a}_y + \cos \theta \hat{a}_z)|_P$$

$$= 6(\sin 0.36\pi \cos 1.7\pi \hat{a}_x + \sin 0.36\pi \sin 1.7\pi \hat{a}_y + \cos 0.36\pi \hat{a}_z)$$

$$= 3.191\hat{a}_x - 4.3921\hat{a}_y + 2.5547\hat{a}_z$$

$$\vec{I}_{\text{Cart}} = \underline{3.191\hat{a}_x - 4.3921\hat{a}_y + 2.5547\hat{a}_z}$$

c) the smallest angle (degrees) between $\vec{H}$ and $\vec{I}$ at point P , and

Use (1.15), $\vec{H} \cdot \vec{I} = |\vec{H}| |\vec{I}| \cos \theta_{HI}$ where

$$\begin{aligned} \vec{H} \cdot \vec{I} &= (-3\hat{a}_x + 5\hat{a}_y - 7\hat{a}_z) \cdot (3.191\hat{a}_x - 4.3921\hat{a}_y + 2.5547\hat{a}_z) \\ &= -3(3.191) + 5(-4.3921) - 7(2.5547) = -49.4165 \end{aligned}$$

$$|\vec{H}| = \sqrt{(-3)^2 + 5^2 + (-7)^2} = \sqrt{83} = 9.1104, \text{ and}$$

$$|\vec{I}| = \sqrt{(3.191)^2 + (-4.3921)^2 + (2.5547)^2} = 6.$$

$$\text{Next, } \cos \theta_{HI} = \frac{\vec{H} \cdot \vec{I}}{|\vec{H}| |\vec{I}|} \Rightarrow \theta_{HI} = \cos^{-1} \left(\frac{\vec{H} \cdot \vec{I}}{|\vec{H}| |\vec{I}|} \right) = \cos^{-1} \left(\frac{-49.4165}{9.1104(6)} \right)$$

$$\theta_P = \underline{154.693^\circ}$$

d) the **vector** component of $\vec{I}$ in the direction of $\vec{H}$ at point P (express in Cartesian coordinates)

Use (1.35), $\vec{I}_H = (\vec{I} \cdot \hat{a}_H) \hat{a}_H$ where

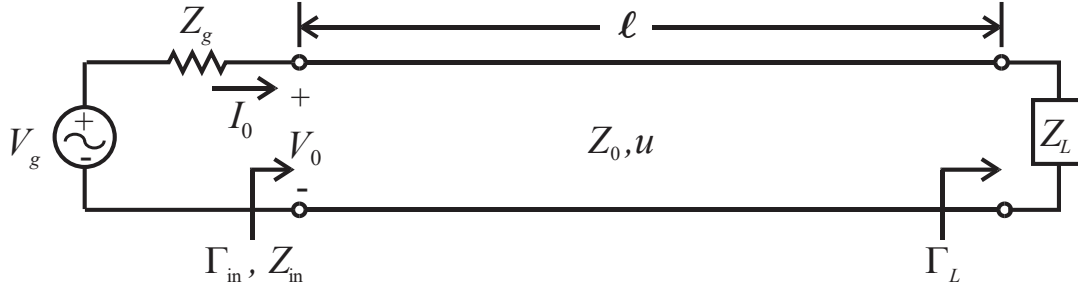
$$\hat{a}_H = \frac{\vec{H}}{|\vec{H}|} = \frac{-3\hat{a}_x + 5\hat{a}_y - 7\hat{a}_z}{9.1104} = -0.329293\hat{a}_x + 0.548821\hat{a}_y - 0.76835\hat{a}_z$$

$$\begin{aligned} \vec{I} \cdot \hat{a}_H &= (3.191\hat{a}_x - 4.3921\hat{a}_y + 2.5547\hat{a}_z) \cdot (-0.329293\hat{a}_x + 0.548821\hat{a}_y - 0.76835\hat{a}_z) \\ &= 3.191(-0.329293) - 4.3921(0.548821) + 2.5547(-0.76835) = -5.4241 \end{aligned}$$

$$\vec{I}_H = (-5.4241)(-0.329293\hat{a}_x + 0.548821\hat{a}_y - 0.76835\hat{a}_z)$$

$$\vec{I}_H = \underline{1.786\hat{a}_x - 2.977\hat{a}_y + 4.168\hat{a}_z}$$

- 3) For the lossless transmission (TL) circuit shown, $Z_0 = 75 \Omega$, $u = 2.25 \times 10^8 \text{ m/s}$, $\ell = 27.9 \text{ cm}$, $f = 2.1 \text{ GHz}$, $Z_L = 30 + j20 \Omega$, $Z_g = 50 \Omega$, and $V_g = 100 \angle 0^\circ \text{ (V)}$. Calculate the wavelength and electrical length (in radians) of the TL. Next, find the load reflection coefficient Γ_L , input impedance, input phasor current & voltage, and time-average power delivered to the load. **Express phasor current & voltage and Γ as $A \angle \theta$ (w/ angle in degrees) and impedance in rectangular format (e.g., $a + jb$).**



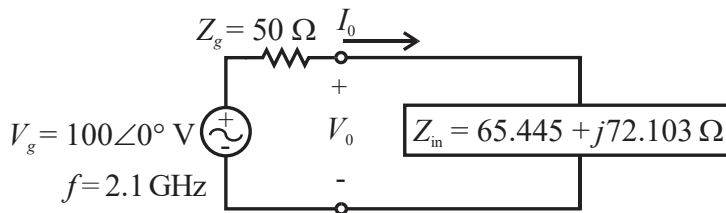
$$\lambda = u/f = 2.25 \times 10^8 / 2.1 \times 10^9 \Rightarrow \underline{\lambda = 0.10714 \text{ m} = 10.714 \text{ cm}}$$

$$\beta \ell = (2\pi/\lambda) \ell = (2\pi/0.10714) 0.279 \Rightarrow \underline{\beta \ell = 16.3614 \text{ rad/m}}$$

$$\Gamma_L = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{(30 + j20) - 75}{(30 + j20) + 75} \Rightarrow \underline{\Gamma_L = 0.4607 \angle 145.253^\circ}$$

$$\Gamma_{in} = \Gamma_L e^{-j2\beta \ell} = (0.4607 \angle 145.253^\circ) e^{-j2(16.3614)} \Rightarrow \underline{\Gamma_{in} = 0.4607 \angle 70.373^\circ}$$

$$Z_{in} = Z_0 \frac{1 + \Gamma_{in}}{1 - \Gamma_{in}} = 75 \left[\frac{1 + (0.4607 \angle 70.373^\circ)}{1 - (0.4607 \angle 70.373^\circ)} \right] \Rightarrow \underline{Z_{in} = 65.445 + j72.103 \Omega}$$



$$I_0 = \frac{V_g}{Z_g + Z_{in}} = \frac{100 \angle 0^\circ}{50 + (65.445 + j72.103)} \Rightarrow \underline{I_0 = 0.73469 \angle -31.987^\circ \text{ A}}$$

$$V_0 = V_g \frac{Z_{in}}{Z_g + Z_{in}} = 100 \angle 0^\circ \frac{65.445 + j72.103}{50 + (65.445 + j72.103)} \Rightarrow \underline{V_0 = 71.5405 \angle 15.784^\circ \text{ V}}$$

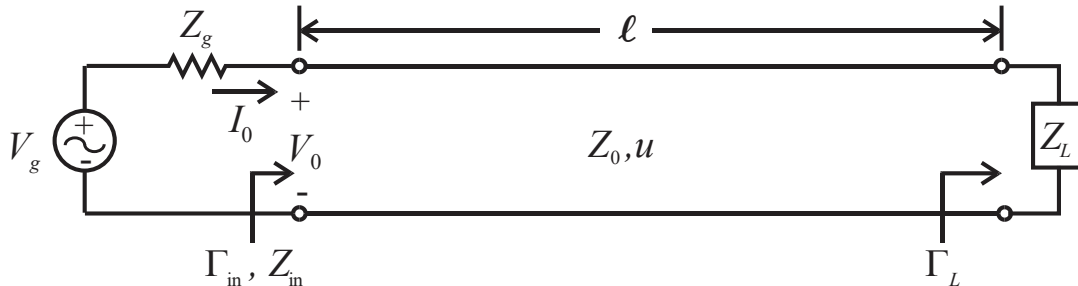
For a lossless TL, the input power and load power are the same.

$$P_L = P_{in} = 0.5 \text{ Re} \{ V_0 I_0^* \} = 0.5 \text{ Re} \{ (71.54 \angle 15.78^\circ) (0.7347 \angle 31.99^\circ) \} \Rightarrow \underline{P_L = 17.663 \text{ W.}}$$

$$\lambda = \underline{10.71 \text{ cm}} \quad \beta \ell = \underline{16.361 \text{ rad/m}} \quad \Gamma_L = \underline{0.4607 \angle 145.253^\circ} \quad Z_{in} = \underline{65.445 + j72.103 \Omega}$$

$$I_0 = \underline{0.7347 \angle -31.99^\circ \text{ A}} \quad V_0 = \underline{71.54 \angle 15.78^\circ \text{ V}} \quad P_L = \underline{17.663 \text{ W}}$$

- 4) For the lossless transmission circuit shown, $Z_0 = 50 \Omega$, $u = 2.4 \times 10^8 \text{ m/s}$, $\ell = 2.7\lambda$, $f = 980 \text{ MHz}$, $Z_{\text{in}} = 130 - j120 \Omega$, $Z_g = 75 \Omega$, and $V_g = 40 \angle 30^\circ \text{ (V)}$. Using a **Smith chart (show & label all work)**, find the input and load reflection coefficients, input admittance, load impedance, and standing wave ratio. Express reflection coefficients in polar form w/ angle in degrees and admittances & impedances in rectangular format.



- Find normalized input impedance $z_{\text{in}} = Z_{\text{in}}/Z_0 = (130 - j120)/50 = \mathbf{2.6 - j2.4 \Omega/\Omega}$. Plot z_{in} .
- Use compass to draw circle centered on Smith chart through z_{in} and make marks on 'REFL. COEFF, Vor I' and 'SWR (VSWR)' scales. Read $|\Gamma_{\text{in}}| = \mathbf{0.67}$ and $\mathbf{SWR = 5}$.
- Use straight edge to draw radial line from center of Smith chart to outer rings. On the 'ANGLE OF REFLECTION COEFFICIENT' scale, read $\angle \Gamma_{\text{in}} = \mathbf{-22.6^\circ}$. Put together to get $\Rightarrow \mathbf{\Gamma_{\text{in}} = 0.67 \angle -22.6^\circ}$.
- Calculate TL length $\ell/\lambda = 2.7\lambda / \lambda = 2.7 \Rightarrow 0.2$. Move 0.2λ on an arc/circle of constant radius $|\Gamma|$ in the 'WAVELENGTHS TOWARD LOAD' direction (CCW) from z_{in} to z_L &/or Γ_L point. Read $\mathbf{z_L = 0.26 + j0.53 \Omega/\Omega}$ and $\angle \Gamma_L = \mathbf{121.4^\circ} \Rightarrow \mathbf{\Gamma_L = 0.67 \angle 121.4^\circ}$.
- Calculate load impedance as $Z_L = z_L Z_0 = (0.26 + j0.53) 50 \Rightarrow \mathbf{Z_L = 13 + j26.5 \Omega}$.
- Move 180° on the circle of constant radius $|\Gamma|$ from z_{in} to y_{in} . Read the normalized input admittance $\mathbf{y_{\text{in}} = 0.208 + j0.193 \text{ S/S}}$.
- Calculate $Y_{\text{in}} = y_{\text{in}} Y_0 = (0.208 + j0.193)/50 = \Rightarrow \mathbf{Y_{\text{in}} = 0.00416 + j 0.00386 \text{ S}}$.

$$\Gamma_{\text{in}} = \mathbf{0.67 \angle -22.6^\circ}$$

$$\Gamma_L = \mathbf{0.67 \angle 121.4^\circ}$$

$$Y_{\text{in}} = \mathbf{4.16 + j 3.86 \text{ mS}}$$

$$Z_L = \mathbf{13 + j26.5 \Omega}$$

$$\text{SWR} = \mathbf{5}$$

Simple Smith Chart

$Z_0 = 50 \, \Omega$
 $f = 980 \text{ MHz}$
 $u = 2.4 \times 10^8 \text{ m/s}$

