

EE 313 Signals and Systems (Fall 2024) Examination 3

Name Key A**Instructions:** Show all work for full credit. Write answers in indicated places. Insert equation sheet inside exam.

- 1) A voltage $v_i(t) = 16 + 7\cos(\omega_0 t) + 6.4\cos(2\omega_0 t)$ (V) $-\infty < t < \infty$ is input into a circuit with a frequency response $H(\omega) = \frac{V_o(\omega)}{V_i(\omega)} = \frac{R_L}{R_s + R_L + 1/j\omega C}$. Find the output $v_o(t)$ voltage when $\omega_0 = 100\pi$ rad/s, $R_s = 50\ \Omega$, $R_L = 150\ \Omega$, and $C = 20\ \mu\text{F}$. As part of the solution process evaluate $H(\omega)$ at $\omega = 0$, ω_0 , and $2\omega_0$ (polar form w/ angle in radians).

$$H(\omega) = \frac{j\omega C R_L}{(R_s + R_L)j\omega C + 1} = \frac{j\omega 150(20 \times 10^{-6})}{(50 + 150)j\omega 20 \times 10^{-6} + 1} = \frac{j\omega 0.003}{1 + j\omega 0.004}$$

$$H(0) = 0$$

$$H(100\pi) = \frac{j 100\pi (0.003)}{1 + j 100\pi (0.004)} = \underline{0.58686 \angle 0.67216}$$

$$H(200\pi) = \frac{j 200\pi (0.003)}{1 + j 200\pi (0.004)} = \underline{0.69686 \angle 0.37868}$$

Per (5.11) $y(t) = A |H(\omega_0)| \cos(\omega_0 t + \theta + \angle H(\omega_0))$.

Therefore,

$$v_o(t) = 16(0) + 7(0.58686) \cos(100\pi t + 0.67216) + 6.4(0.69686) \cos(200\pi t + 0.37868) \checkmark$$

$$= 4.108 \cos(100\pi t + 0.67216) + 4.46 \cos(200\pi t + 0.3787) \checkmark$$

$$H(0) = \underline{0} \quad H(\omega_0) = \underline{0.587 \angle 0.672} \quad H(2\omega_0) = \underline{0.697 \angle 0.379}$$

$$v_o(t) = \underline{4.108 \cos(100\pi t + 0.6722) + 4.46 \cos(200\pi t + 0.3787) \checkmark} \quad -\infty < t < \infty$$

- 2) The transfer function of a linear, time-invariant, casual (no initial energy), discrete-time filter is given by

$$H(z) = \frac{z^2 + 0.2z + 0.27}{z^2 + 0.36}$$

First, find the partial fraction expansion of $H(z)/z$. Then, find an entirely real, analytic expression for the unit-pulse response $h[n]$ of the system for all $n \geq 0$. Calculate $h[0]$, $h[1]$, and $h[3]$.

$$z^2 + 0.36 = 0 \Rightarrow z = \sqrt{-0.36} \Rightarrow z = \pm j0.6$$

$$\frac{H(z)}{z} = \frac{z^2 + 0.2z + 0.27}{z(z - j0.6)(z + j0.6)} = \frac{C_0}{z} + \frac{C_1}{z - j0.6} + \frac{C_2}{z + j0.6}$$

$$\uparrow p_1 = j0.6 = 0.6 \angle \frac{\pi}{2}$$

Residues $C_0 = H(0) = \frac{0.27}{0.36} = 0.75$

$$C_1 = \left\{ (z - j0.6) \frac{z^2 + 0.2z + 0.27}{z(z - j0.6)(z + j0.6)} \right\} \bigg|_{z=j0.6} = \frac{(j0.6)^2 + 0.2(j0.6) + 0.27}{j0.6(j0.6 + j0.6)}$$

$$= \frac{0.2083 \angle -0.9273}{1} = 0.2083 \angle -0.9273$$

$$C_2 = C_1^* = 0.2083 \angle +0.9273 = 0.2083 \angle 0.9273$$

$$\frac{H(z)}{z} = \frac{0.75}{z} + \frac{0.2083 \angle -0.9273}{z - j0.6} + \frac{0.2083 \angle 0.9273}{z + j0.6}$$

$$H(z) = 0.75 + \frac{0.2083 \angle -0.9273 z}{z - j0.6} + \frac{0.2083 \angle 0.9273 z}{z + j0.6}$$

Use linearity, Table 7.3 $\delta[n] \leftrightarrow 1$, and (7.68)

$2|C_1| \sigma^n \cos(\omega n + \phi_1)$ where $\sigma = |p_1| = 0.6$ & $\omega = \angle p_1 = +\frac{\pi}{2}$ $n \geq 0$

$$h[n] = 0.75 \delta[n] + 2(0.2083) 0.6^n \cos\left(+\frac{\pi}{2} n - 0.9273\right) u[n]$$

$$h[n] = 0.75 \delta[n] + 0.416 (0.6)^n \cos\left(\frac{\pi}{2} n - 0.9273\right) u[n]$$

$$h[0] = 0.75 + 0.416 \cos(-0.9273) = 1 \quad h[3] = 0.416 (0.6)^3 \cos(1.5\pi - 0.93) = -0.072$$

$$h[1] = 0.416 (0.6) \cos(\frac{\pi}{2} - 0.9273) = 0.2$$

$$H(z)/z = \frac{0.75}{z} + \frac{0.2083 \angle -0.9273}{z - j0.6} + \frac{0.2083 \angle 0.9273}{z + j0.6}$$

$$h[n] = 0.75 \delta[n] + 0.416 (0.6)^n \cos\left(\frac{\pi}{2} n - 0.9273\right) u[n]$$

$$h[0] = 1 \quad h[1] = 0.2 \quad h[3] = -0.072$$

- 3) Given signal $x(t) = 8 + 4\sin(300t)$, $-\infty < t < \infty$, find the frequency spectrum $X(\omega)$ and draw a fully labeled sketch of the magnitude of $|X(\omega)|$ on provided axes. If $x(t)$ is to be sampled, determine the maximum sampling rate $T_{\max}$ and corresponding minimum sampling frequency $\omega_{\min}$ to avoid aliasing. If $x(t)$ is then sampled at a rate $T = \pi/250$ s = 12.5664 ms, does the sampled signal $x[n] = x(nT)$ experience aliasing? For $x[n]$, draw a fully labeled sketch of the magnitude of the frequency spectrum $|X_s(\omega)|$ on the provided axis for the frequency range shown.

Using linearity + Table 3.2 $1 \leftrightarrow 2\pi f(\omega)$

$$\sin(\omega_0 t) \leftrightarrow j\pi [\delta(\omega + \omega_0) - \delta(\omega - \omega_0)]$$

$$X(\omega) = 8(2\pi f(\omega)) + 4j\pi[f(\omega+300) - f(\omega-300)]$$

From $|X(\omega)|$, $B = 300 \frac{\text{rad}}{\text{s}} \Rightarrow \omega_{\min} = 2B = \underline{600 \frac{\text{rad}}{\text{s}}}$

$$T_{\max} = \frac{2\pi}{\omega_{\min}} = \frac{2\pi}{600} = \frac{\pi}{300} = 0.010472 \text{ s} = \underline{10.472 \text{ ms}}$$

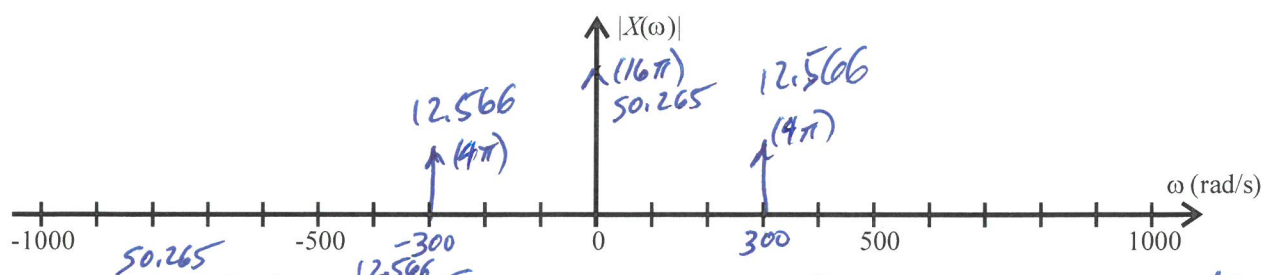
Since $T = 12.5664 \text{ ms} > T_{\max} \Rightarrow \underline{\text{aliasing!}}$

Per (5.51), $X_5(\omega) = \sum_{k=-\infty}^{\infty} \frac{1}{T} X(\omega - k\omega_s)$

$$T = \pi/250 \quad \omega_s = \frac{2\pi}{T} = 500 \frac{\text{rad}}{\text{s}}$$

$$\frac{1}{T} = \frac{250}{\pi}$$

$$16\pi\left(\frac{250}{\pi}\right) = 4000 \quad + \quad 4\pi\left(\frac{250}{\pi}\right) = 1000$$

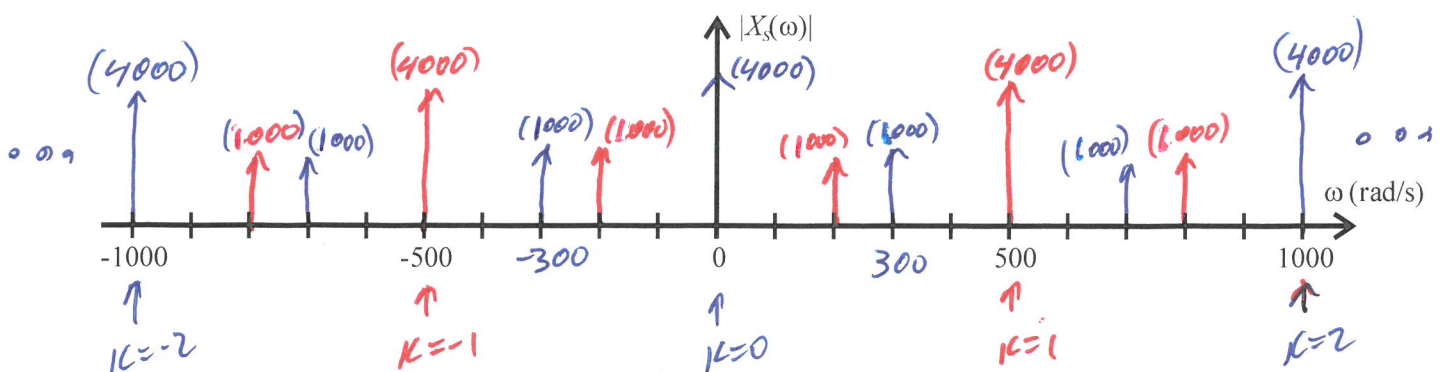


$$X(\omega) = \frac{16\pi f(\omega) + j4\pi [f(\omega+300) - f(\omega-300)]}{2.506}$$

$$\omega_{\min} = 600 \text{ rad/s}$$

$$T_{\max} = 10.472 \text{ ms}$$

If $T = 12.5664$ ms, is there aliasing? YES / NO (circle correct answer)



- 4) The transfer function of a linear, time-invariant, casual (no initial energy), discrete-time filter is given by

$$H(z) = \frac{z + 0.5}{z^2 - 0.1z + 0.72}.$$

- a) For an input $x[n] = 6u[n]$, find $X(z)$ and the output $Y(z)$ of the system as a rational function.

Use linearity and Table 7.3 $u[n] \leftrightarrow \frac{z}{z-1}$, to get

$$\underline{X(z) = \frac{6z}{z-1}}. \text{ Per Table 7.2 convolution property}$$

$$Y(z) = X(z)H(z) = \frac{6z(z+0.5)}{(z-1)(z^2-0.1z+0.72)} = \frac{6z^2+3z}{(z^3-0.1z^2+0.72z)-(z^2-0.1z+0.72)}$$

=

$$X(z) = \underline{\frac{6z}{z-1}} \quad Y(z) = \underline{\frac{6z^2+3z}{z^3-1.1z^2+0.82z-0.72}}$$

- b) Without solving for $y[n]$, determine $y[0]$ and $y[1]$

Per Initial-Value Theorem (Table 7.2)

$$X[0] = \lim_{z \rightarrow \infty} X(z) \Rightarrow y[0] = \lim_{z \rightarrow \infty} \frac{6/z + 3/z^2}{1 - \dots} = \underline{\underline{0}}$$

$$X[1] = \lim_{z \rightarrow \infty} [zX(z) - zX[0]] \Rightarrow y[1] = \lim_{z \rightarrow \infty} \frac{6 + 3/z}{1 - \dots} = \underline{\underline{6}}$$

$$y[0] = \underline{0} \quad y[1] = \underline{6}$$

- c) Without solving for $y[n]$, determine $y[n \rightarrow \infty]$.

$$z^2 - 0.1z + 0.72 \Rightarrow \text{poles are } z = 0.8485 \pm 1.5118j \quad |z| < 1$$

Final-value Theorem Table 7.3

$$\lim_{n \rightarrow \infty} y[n] = (z-1)Y(z) \Big|_{z=1} = \frac{(\cancel{z-1})(6z^2+3z)}{(\cancel{z-1})(z^2-0.1z+0.72)}$$

$$= \frac{6(1)^2+3(1)}{1^2-0.1(1)+0.72} = \frac{9}{1.62}$$

$$y[n \rightarrow \infty] = \underline{5.55}$$